MARC4 – 4-bit Universal Microcontroller

The M44C510 is a member of the TEMIC family of 4-bit single chip microcontrollers. It contains ROM, RAM, up to 30 digital I/O pins, up to 10 maskable external interrupt sources, 6 maskable internal interrupts, a watchdog timer, 32-kHz oscillator with programmable watch timer, 2 x 8-bit multifunction timer/counter module and a versatile on-chip system clock generation module.

Features

- 4K x 8-bit application ROM
- 256 x 4-bit RAM
- 8 hardware and software interrupt priority levels
- Bitwise maskable prioritised interrupts
- Up to 10 external and 4 internal interrupts
- Up to 30 I/O lines
- High drive ports (20 mA, $V_{DD} = 5 \text{ V}$)
- I/O ports bitwise configurable with combined interrupt handling (for serial I/O applications)
- 2 x 8-bit multifunction timer/counters
- 32-kHz on chip oscillator with programmable watch timer
- User definable on-chip system clock generation
- 4-MHz crystal, 4-MHz ceramic resonator or fully intergrated RC oscillator **

Benefits

- Extremely low power consumption
- Minimal external components
- Coded reset and watchdog timer **
- Power on reset, "brown out" function
- Power down mode
- 2.4 V to 6.2 V supply voltage
- Data retention down to 2.2 V in SLEEP mode
- Efficient, hardware controlled interrupt handling
- High level programming language in qFORTH
- Comprehensive library of useful routines
- PC based development tools

(** mask option)

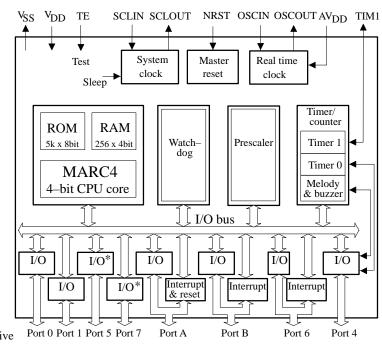


Figure 1. Block diagram

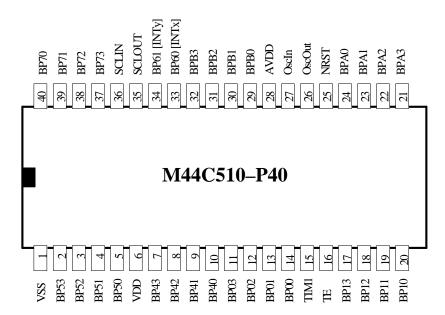


Figure 2. Pin connections

Table 1. Pin description

Name	Function
$V_{ m DD}$	Power supply voltage +2.4 V to +6.2 V
AV_{DD}	Analogue power supply voltage +2.4 V to +6.2V
V_{SS}	Circuit ground
BP00 – BP03	4 bidirectional I/O lines of Port 0 – automatic nibblewise configurable I/O
BP10 – BP13	4 bidirectional I/O lines of Port 1(*) – automatic nibblewise configurable I/O
BP50 – BP53	4 bidirectional I/O lines of high current Port 5(*) – bitwise configurable I/O
BP70 – BP73	4 bidirectional I/O lines of high current Port 7(*) – bitwise configurable I/O
BPA0 – BPA3	4 bidirectional I/O lines of Port A(*) – bitwise configurable I/O and as inputs to a port monitor
	module. Optional coded reset inputs (*)
BPB0 – BPB3	4 bidirectional I/O lines of Port B(*) – bitwise configurable I/O and as inputs to a port monitor
	module
BP60 – BP61	2 bidirectional I/O lines of Port 6 (*) – bitwise configurable I/O or as 2 external programmable
	interrupts
BP40-T0OUT0	I/O line BP40 of Port 4(*) – configurable I/O or timer/counter 0 I/O T0OUT0
BP41-T0OUT1	I/O line BP41 of Port 4(*) – configurable I/O or timer/counter 0 I/O T0OUT1
BP42-BUZ	High current I/O line BP42 of Port 4(*) – configurable I/O or buzzer output BUZ
BP43-NBUZ	High current I/O line BP43 of Port 4(*) – configurable I/O or buzzer output NBUZ
TIM1	Dedicated bidirectional I/O for Timer 1
SCLIN	4-MHz quartz crystal/ceramic resonator or trimming resistor pin (mask option dependent)
SCLOUT	4-MHz quartz crystal/ceramic resonator pin (mask option dependent)
OSCIN	32-kHz quartz crystal pin (mask option dependent)
OSCOUT	32-kHz quartz crystal pin (mask option dependent)
TE	Testmode input. This input is used to control the test modes (internal pull-down)
NRST	Reset input (/output), a logic low on this pin resets the device. An internal watchdog or coded reset is indicated by a low pulse on this pin.

^(*) For mask options please see the order information.

M44C510

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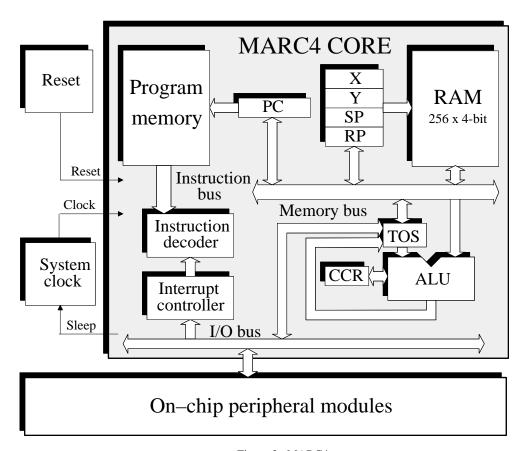
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1 MARC4 Architecture

1.1 General Description

The MARC4 microcontroller consists of an advanced stack based 4-bit CPU core and on-chip peripherals. The CPU is based on the HARVARD architecture with physically separate program memory (ROM) and data memory (RAM). Three independent buses, the instruction bus, the memory bus and the I/O bus are used for parallel communication between ROM, RAM and peripherals. This enhances program execution speed by allowing both

instruction prefetching, and a simultaneous communication to the on-chip peripheral circuitry. The extremely powerful integrated interrupt controller with associated eight prioritized interrupt levels supports fast and efficient processing of hardware events. The MARC4 is designed for the high level programming language qFORTH. The core includes an expression and a return stack. This architecture allows high level language programming without any loss in efficiency or code density.



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Figure 3. MARC4 core

1.2 Components of MARC4 Core

The core contains ROM, RAM, ALU, program counter, RAM address registers, instruction decoder and interrupt controller. The following sections describe each functional block in more detail:

1.2.1 ROM

The program memory (ROM) is mask programmed with the customer application program during the fabrication of the microcontroller. The ROM is addressed by a 12-bit wide program counter, thus predefining a maximum program bank size of 4 Kbytes. An additional 1 Kbyte of ROM exists which is accessible using a ROM bank switch. Of this program space a section is reserved for quality control self-test software, the remainder is available for application program. The lowest user ROM address segment is taken up by a 512 byte zero page which contains predefined start addresses for interrupt service routines and special subroutines accessible with single byte instructions (SCALL). The corresponding memory map is shown in figure 4. Look-up tables of constants can also be held in ROM and are accessed via the MARC4's built-in TABLE instruction.

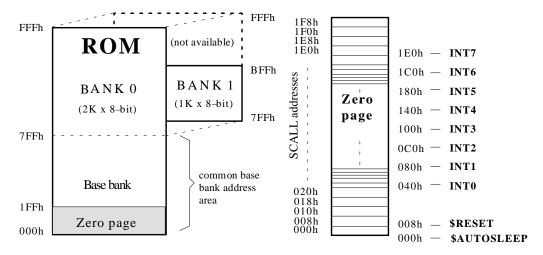


Figure 4. ROM map of M44C510

ROM Banking

For customers programming with qFORTH the bank switching is fully supported by the compiler. The MARC4 switches from one ROM bank to another by writing the new bank number to the ROM Bank Register (RBR). Conventional program space (power up bank) resides in ROM bank 0. Each ROM bank consists of a 4 Kbyte address space whereby the lowest 2 Kbyte is common to all banks, so that addresses between 000h and 7FFh always accesses the same ROM data (see figure 4). When ROM banking is used, the compiler will, if necessary insert program code to save and restore the condition of the RBR on bank switching.

1.2.2 RAM

The MARC4 contains 256 x 4-bit wide static random access memory (RAM). It is used for the expression stack, the return stack and data memory for variables and arrays. The RAM is addressed by any of the four 8-bit wide RAM address registers SP, RP, X and Y.

Expression Stack

The 4-bit wide expression stack is addressed with the expression stack pointer (SP). All arithmetic, I/O and memory reference operations take their operands from, and return their result to the expression stack. The MARC4 performs the operations with the top of stack items (TOS and TOS-1). The TOS register contains the top element of the expression stack and works like an accumulator. This stack is also used for passing parameters between subroutines, and as a scratch pad area for temporary storage of data.

Return Stack

The 12-bit wide return stack is addressed by the return stack pointer (RP). It is used for storing return addresses of subroutines, interrupt routines and for keeping loop index counts. The return stack can also be used as a temporary storage area.

The MARC4 instruction set supports the exchange of data between the top elements of the expression stack and the return stack. The two stacks within the RAM have a user definable location and maximum depth.

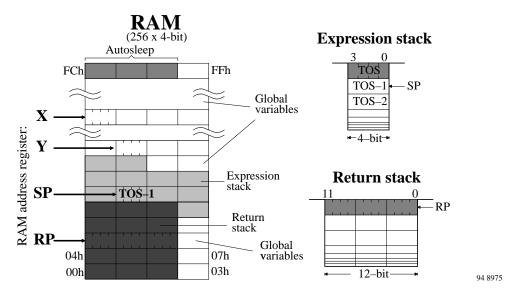


Figure 5. RAM map

1.2.3 Registers

The MARC4 controller has seven programmable registers and one condition code register. They are shown in the following programming model.

Program Counter (PC)

The program counter (PC) is a 12-bit register that contains the address of the next instruction to be fetched from the ROM. Instructions currently being executed are decoded in the instruction decoder to determine the internal micro operations. For linear code (no calls or branches) the program counter is incremented with every instruction cycle. If a branch-, call-, return-instruction or an interrupt is executed the program counter is loaded with a new address. The program counter is also used with the TABLE instruction to fetch 8-bit wide ROM constants.

ROM Banking Register (RBR)

The ROM banking register is a 4-bit register whereby in the M44C510 only bit 2 is used. This indicates which ROM bank is presently being addressed. The RBR is accessed with a standard qFORTH peripheral read or write instruction (IN or OUT, port address 'D' hex).

RAM Address Registers

The RAM is addressed with the four 8-bit wide RAM address registers: SP, RP, X and Y. These registers allow access to any of the 256 RAM nibbles.

Expression Stack Pointer (SP)

The stack pointer (SP) contains the address of the next-totop 4-bit item (TOS-1) of the expression stack. The pointer is automatically pre-incremented if a nibble is moved onto the stack or post-decremented if a nibble is removed from the stack. Every post-decrement operation moves the item (TOS-1) to the TOS register before the SP is decremented. After a reset the stack pointer has to be initialized with ">SP SO" to allocate the start address of the expression stack area.

Return Stack Pointer (RP)

The return stack pointer points to the top element of the 12-bit wide return stack. The pointer automatically preincrements if an element is moved onto the stack or it post-decrements if an element is removed from the stack. The return stack pointer increments and decrements in steps of 4. This means that every time a 12-bit element is stacked, a 4-bit RAM location are left unwritten. These location are used by the qFORTH compiler to allocate 4-bit variables. After a reset the return stack pointer has to be initialized with ">RP FCh".

RAM Address Register (X and Y)

The X and Y registers are used to address any 4-bit item in the RAM. A fetch operation moves the addressed nibble onto the TOS. A store operation moves the TOS to the addressed RAM location. Using either the pre-increment or post-decrement addressing mode arrays in the RAM can be compared, filled or moved.

Top Of Stack (TOS)

The top of stack register is the accumulator of the MARC4. All arithmetic/logic, memory reference and I/O

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operations use this register. The TOS register receives data from the ALU, ROM, RAM or I/O bus.

Condition Code Register (CCR)

The 4-bit wide condition code register contains the branch, the carry and the interrupt enable flag. These bits indicates the current state of the CPU. The CCR flags are set or reset by ALU operations. The instructions SET_BCF, TOG_BF, CCR! and DI allow a direct manipulation of the condition code register.

Carry/Borrow (C)

The carry/borrow flag indicates that borrow or carry out of arithmetic logic unit (ALU) occurred during the last arithmetic operation. During shift and rotate operations this bit is used as a fifth bit. Boolean operations have no affect on the C flag.

Branch (B)

The branch flag controls the conditional program branching. Should the branch flag have been set by a previous instruction a conditional branch will cause a jump. This flag is affected by arithmetic, logic, shift, and rotate operations.

Interrupt Enable (I)

The interrupt enable flag globally enables or disables the triggering of all interrupt routines with the exception of the non-maskable reset. After a reset or on executing the DI instruction the interrupt enable flag is reset thus disabling all interrupts. The core will not accept any further interrupt requests until the interrupt enable flag has been set again by either executing an EI, RTI or SLEEP instruction.

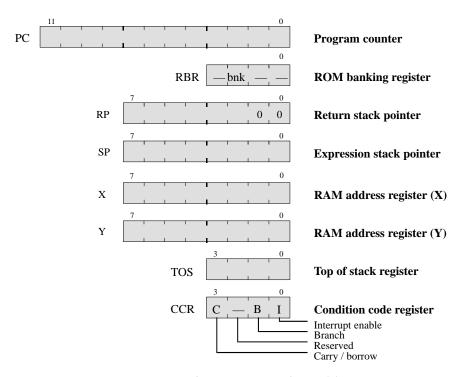


Figure 6. Programming model

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1.2.4 ALU

The 4-bit ALU performs all the arithmetic, logical, shift and rotate operations with the top two elements of the expression stack (TOS and TOS-1) and returning the result to the TOS. The ALU operations affect the carry/borrow and branch flag in the condition code register (CCR).

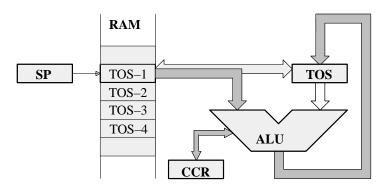


Figure 7. ALU zero address operations

1.2.5 Instruction Set

The MARC4 instruction set is optimized for the high level programming language qFORTH. Many MARC4 instructions are qFORTH words. This enables the compiler to generate fast and compact program code. The CPU has an instruction pipeline allowing the controller to prefetch instruction from ROM at the same time as the present instruction is being executed. The MARC4 is a zero address machine, the instructions containing only the operation to be performed and no source or destination address fields. The operations are implicitly performed on the data placed on the stack. There are one and two byte instructions which are executed within 1 to 4 machine cycles. A MARC4 machine cycle is made up of two system clock (SYSCL) cycles. Most of the instructions are only one byte long and are executed in a single

machine cycle. For more information see section "MARC4 instruction set overview".

1.2.6 I/O Bus

The I/O ports and the registers of the peripheral modules (Timer 0, Timer 1, Watch timer, Watchdog etc.) are I/O mapped. All communication between the core and the onchip peripherals takes place via the I/O bus and the associated I/O control. With the MARC4 IN and OUT instructions the I/O bus allows a direct read or write access to one of the 16 primary I/O addresses. More about the I/O access to the on-chip peripherals is described in the section "Peripheral Modules". The I/O bus is internal and is not accessible by the customer on the final microcontroller device, but it is used as the interface for the MARC4 emulation (see also the section "Emulation").

1.3 Interrupt Structure

The MARC4 can handle interrupts with eight different priority levels. They can be generated from the internal and external interrupt sources or by a software interrupt from the CPU itself. Each interrupt level has a hard-wired priority and an associated vector for the service routine in the ROM (see table 2). The programmer can postpone the processing of interrupts by resetting the interrupt enable

flag (I) in the CCR. An interrupt occurrence will still be registered but the interrupt routine only started after the I flag is set. All interrupts can be masked, and the priority individually software configured by programming the appropriate control register of the interrupting module (see section "Peripheral Modules").

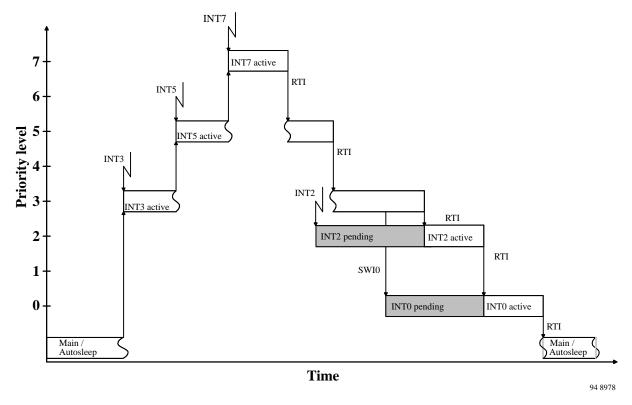


Figure 8. Interrupt handling

Interrupt Processing

For processing the eight interrupt levels the MARC4 includes an interrupt controller with two 8-bit wide "interrupt pending" and "interrupt active" registers. The interrupt controller samples all interrupt requests during every non-I/O instruction cycle and latches these in the interrupt pending register. If no higher priority interrupt is present in the interrupt active register it signals the CPU to interrupt the current program execution. If the interrupt enable bit is set the processor enters an interrupt acknowledge cycle. During this cycle a short call (SCALL) instruction to the service routine is executed and the current PC is saved on the return stack. An interrupt service routine is finished with the RTI instruction. This instruction sets the interrupt enable flag, resets the corresponding bits in the interrupt pending/active register and fetches the return address from the return stack to the program counter. When the interrupt enable flag is reset (triggering of interrupt routines are disabled), the execution of new interrupt service routines is inhibited but not the logging of the interrupt requests in the interrupt pending register. The execution of the interrupt will be delayed until the interrupt enable flag is set again. Note that interrupts are only lost if an interrupt request occurs while the corresponding bit in the pending register is still set (i.e. the interrupt service routine is not yet finished).

It should also be realised that automatic stacking of the RBR is not carried out by the hardware and so if ROM banking is used, the RBR must be stacked on the expression stack by the application program and restored before the RTI. After a master reset (power-on, external or watchdog reset), the interrupt enable flag and the interrupt pending and interrupt active register are all reset.

Interrupt Latency

The interrupt latency is the time from the occurrence of the interrupt to the interrupt service routine being activated. In the MARC4 this is extremely short taking between 3 to 5 machine cycles depending on the state of the core.

Table 2. Interrupt priority table

Interrupt	Priority	ROM Address	Maskable	Interrupt Opcode
INT0	lowest	040h	Yes	C8h (SCALL 040h)
INT1		080h	Yes	D0h (SCALL 080h)
INT2		0C0h	Yes	D8h (SCALL 0C0h)
INT3		100h	Yes	E8h (SCALL 100h)
INT4		140h	Yes	E8h (SCALL 140h)
INT5		180h	Yes	F0h (SCALL 180h)
INT6	\downarrow	1C0h	Yes	F8h (SCALL 1C0h)
INT7	highest	1E0h	Yes	FCh (SCALL 1E0h)

1.3.1 Hardware Interrupts

In the M44C510 there are eleven hardware interrupt sources which can be programmed to occupy a variety of priority levels. Each source can be individually masked by mask bits in the corresponding control registers. An overview of the possible hardware configurations is shown in table 3.

Table 3. Hardware interrupts

Interrupt	F	Possible Interrupt Priorities				S	RST	Interrupt Mask		Function		
Source	0	1	2	3	4	5	6	7		Register	Bit	
NRST external									X	_	ı	low level active
Watchdog									#	_	ı	1/2 - 2 sec. time out
Port A coded reset									#	_	ı	level any inputs
Port A monitor		*		*		*		*		PAIPR	3	any edge any input
Port B monitor		*		*		*		*		PBIPR	3	any edge any input
Port 60 external		*		*		*		*		P6CR	1,0	any edge
Port 61 external	*		*		*		*			P6CR	3,2	any edge
Watch timer INTA		*				*				WTICR	0	1 of 8 frequencies (1 – 128Hz)
Watch timer INTB			*				*			WTICR	1	1 of 8 frequencies (8 – 8192Hz)
Timer 0		*		*		*		*		T0CR	0	overflow/compare/ end measurement
Timer 1	*		*		*		*			T1CR	0	compare

X = hardwired (neither optional or software configurable)

= customer mask option (see "Ordering Information")

1.3.2 Software Interrupts

The programmer can generate interrupts using the software interrupt instruction (SWI) which is supported in qFORTH by predefined macros named SWI0...SWI7. The software triggered interrupt operates exactly like any hardware triggered interrupt.

The SWI instruction takes the top two elements from the expression stack and writes the corresponding bits via the I/O bus to the interrupt pending register. Thus using the SWI instruction, interrupts can be re-prioritised or lower priority processes scheduled for later execution.

^{* =} software configurable (see "Peripheral Modules" section for further details)

1.4 Master Reset

The master reset forces the CPU into a well-defined condition, is unmaskable and is activated independent of the current program state. It can be triggered by either initial supply power-up, a short collapse of the power supply, a watchdog time out, activation of the NRST input or the occurrence of a coded reset on Port A (see figure 9). A master reset activation will reset the interrupt enable flag, the interrupt pending register and the interrupt active register. During the reset phase the I/O bus control signals are set to 'reset mode' thereby initializing all on-chip peripherals.

Releasing the reset results in a short call instruction (opcode C1h) to the ROM address 008h. This activates the initialization routine \$RESET which in turn initialises all necessary RAM variables, stack pointers and peripheral configuration registers.

Power-on Reset

The fully integrated power-on reset circuit ensures that the core is held in a reset state until the minimum operating supply voltage has been reached. A reset condition will also be generated should the supply voltage drop momentarily below the minimum operating supply.

External Reset (NRST)

An external reset can be triggered with the NRST pin. To activate an external reset the pin should be low for a minimum of two machine cycles.

Coded Reset (Port A)

The coded reset circuit is connected directly to the Port A terminals. Using a mask option, the user can define a hardwired code combination (e.g. all pins low), which if occurring on the Port A will generate a reset in the same way as the NRST pin.

Note that if this option is used, the reset is not maskable and will also trigger if the predefined code is written on to the Port A by the CPU itself. Care should also be taken not to generate an unwanted reset by inadvertently passing through the reset code on input transitions. This applies especially if the pins have a high capacitive loading.

Watchdog Reset

The watchdog can be enabled via mask option and triggers a reset with every watchdog counter overflow. To suppress the watchdog reset, the counter must be regularly reset by reading the watchdog register address (WDRES).

The CPU reacts in exactly the same manner to a reset stimulus from any of the above sources.

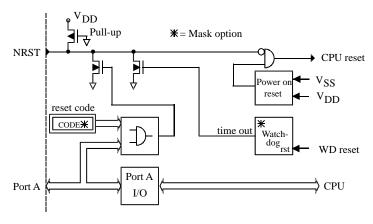


Figure 9. Reset configuration

1.5 Clock Generation

The M44C510 has a dual clock system, a 2-MHz system clock (SYSCL) for the core and a 32-kHz subclock (SUBCL) for the time-keeping peripheral modules (see figure 10). Each clock can be generated from independent on-chip oscillators or they can both be derived from the same high frequency SYSCL oscillator. This is mask selectable - allowing a choice of either a 4-MHz crystal, 4-MHz ceramic resonator or RC oscillators. All the necessary oscillator circuitry, with the exception of the actual crystal or resonator, are integrated on chip. So, for example, if no exact timing is required, it is possible to use the fully integrated RC oscillator, thus operating without any external components.

An additional mask option enables a high resolution trimmable RC oscillator whereby the SYSCL can be trimmed with an external resistor between SCLIN and V_{DD}. In this configuration, the SYSCL frequency can be maintained

stable to within a tolerance of $\pm 10\%$ over the full operating temperature range. For example an RC oscillation frequency of 4 MHz can be obtained by connecting a 2.2 M Ω resistor at 5 Volt and an 1.2 M Ω resistor at 3 Volt supply.

Some applications require only long term time keeping or low resolution timing. In this case, an on-chip low-power 32-kHz crystal oscillator can be used to generate the SUBCL. This allows the core to go into SLEEP mode when not used, thus greatly reducing power consumption.

Should however, the full 2-MHz timing resolution be required, then either the crystal or resonator oscillator should be used for SYSCL generation.

Should a suitable external 4-MHz or 32-kHz clock source be available, then these can be input onto either SCLIN (Crystal oscillator configuration) or OSCIN respectively.

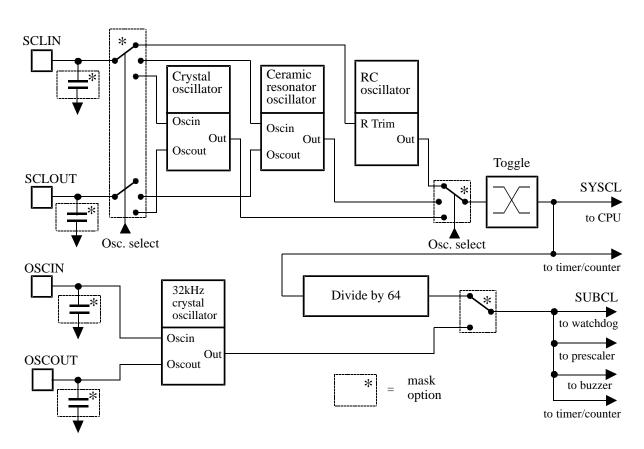


Figure 10. Clock module

Clock Trimming

For trimming purposes, the M44C510 can be put into a clock monitor mode. To do this the test input (TE) must

be pulsed high once, whereupon the SYSCL clock will appear on BP10 (Port 1, bit 0). To put BP10 back into normal operation, the reset must be reapplied (see figure 11).

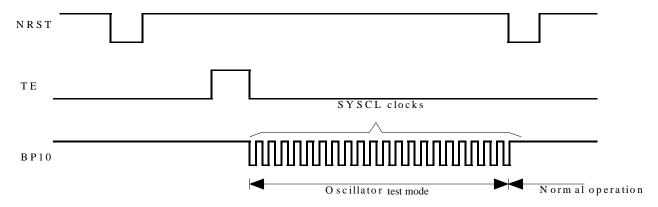


Figure 11. Clock monitoring

1.6 Sleep Mode

The sleep mode is a shutdown condition which is used to reduce the average system power consumption in applications where the μC is not fully utilised. In this mode the system clock is stopped. The sleep mode is entered with the SLEEP instruction. This instruction sets the condition code register interrupt enable bit (I) to enable all interrupts and stops the core. During the sleep mode the peripheral modules remain active and are able to generate interrupts. The μC exits the sleep mode with any interrupt or a reset. The sleep mode can only be maintained when no interrupt pending or active register bits are set. The

application of the \$AUTOSLEEP routine ensures the correct function of the sleep mode. The total power consumption is directly proportional to the active time of the μ C. For a rough estimation of the expected average system current consumption, the following formula should be used:

$$I_{total} (V_{DD}, f_{syscl}) = I_{Sleep} + (I_{DD} * T_{active} / T_{total})$$

 I_{DD} depends on V_{DD} and f_{syscl} . The SLEEP current (I_{Sleep}) is typical less than 1 μA . The active time of the core and the total emulation time are displayed in a separate window of the MARC4 emulator software.

2 Peripheral Modules

2.1 Addressing Peripherals

Accessing the peripheral modules takes place via the I/O bus (see figure 12). The IN or OUT instructions allow direct addressing of up to 16 I/O modules. A dual register addressing scheme has been adopted, with direct addressing of the "primary register". To address the "auxiliary register" the access must be switched with an "auxiliary switching module". Thus a single IN (or OUT) to the module address will read (or write) into the module primary register. Accessing the auxiliary register is

performed with the same instruction preceded by writing the module address into the auxiliary switching module. Byte wide registers are access by multiple IN (or OUT) instructions. For more complex peripheral modules, with a larger number of registers, extended addressing is used. In this case a bank of up to 16 subport registers are indirectly addressed with the subport address being initially written into the auxiliary register.

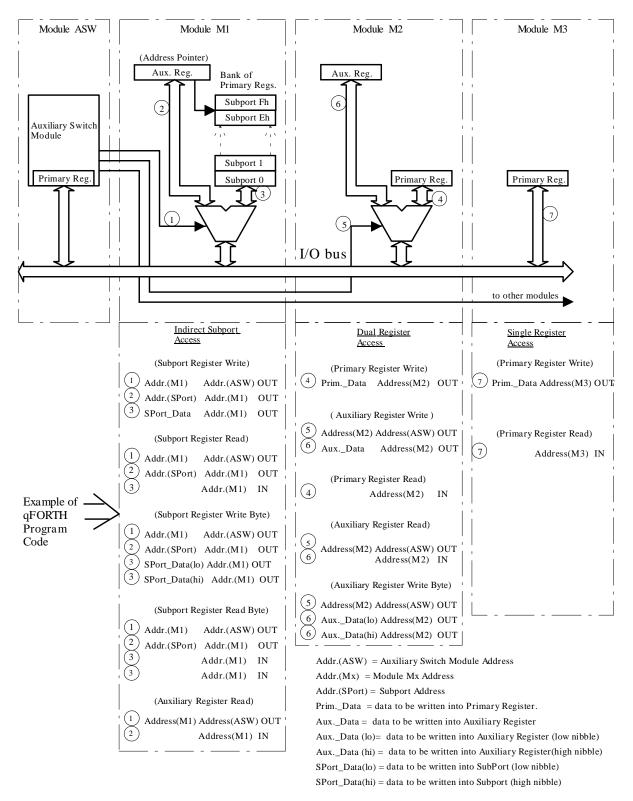


Figure 12. Example of I/O addressing

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Table 4. M44C510 Peripheral addresses

	Address		Name	Write /Read	Function	
	0			P0DAT	W/R	Port 0 – Data Register/Input data
	1			P1DAT	W/R	Port 1 – Data Register/Input data
	2			PAIPR	W	Port A – Interrupt Priority Register
		Αι	IX.	PAICR	W	Port A – Interrupt Control Register
	3			PBIPR	W	Port B – Interrupt Priority Register
				WDRES	R	Watchdog Reset
		Αι	IX.	PBICR	W	Port B – Interrupt Control Register
	4			P4DAT	W/R	Port 4 – Data Register/Pin data
		Αι	ıx.	P4DDR	W	Port 4 – Data Direction Register
	5	•		P5DAT	W/R	Port 5 – Data Register/Pin data
		Αι	IX.	P5DDR	W	Port 5 – Data Direction Register
	6	•		P6DAT	W/R	Port 6 – Data Register/Pin data
		A	ux	P6CR	W	Port 6 – Control Register (Byte)
	7	-		P7DAT	W/R	Port 7 – Data Register/Pin data
		Αι	IX.	P7DDR	W	Port 7 – Data Direction Register
	8	•		ASW	W	Auxiliary Switch Register
9				TCM	W/R	Data to/from Subport addressed by TCX
		Aux.		T0SR	R	Timer 0 Interrupt Status Register
				TCX	W	Timer/Counter Subport Address Pointer
Sub	port Add	lress		•	'	
	-		0	ТОМО	W	Timer 0 Mode Register
			1	T0CR	W	Timer 0 Control Register
			2	T1MO	W	Timer 1 Mode Register
			3	T1CR	W	Timer 1 Control Register
			4	TCMO	W	Timer/Counter Mode Register
			5	TCIO	W	Timer/Counter I/O Control Register
			6	TCCR	W	Timer/Counter Control Register
			7	TCIP	W	Timer/Counter Interrupt Priority
			8	T1CP	W	Timer 1 Compare Register (Byte)
				T1CA	R	Timer 1 Capture Register (Byte)
			9	T0CP	W	Timer 0 Compare Register (Byte)
				T0CA	R	Timer 0 Capture Register (Byte)
			A	BZCR	W	Buzzer Control Register
	B-F		B-F	•	'	Reserved
	A		•	PADAT	W/R	Port A – Data Register/Pin data
		Αι	IX.	PADDR	W	Port A – Data Direction Register
	В			PBDAT	W/R	Port B – Data Register/Pin data
		Αι	IX.	PBDDR	W	Port B – Data Direction Register
	С	•		_	_	Reserved
	D			RBR	W	Rom Bank Switch Register
	Е			_	_	Reserved
	F			WTFSR	W	Watch Timer Frequency Select Register
	F Aux.			1	W	Watch Timer Interrupt Priority Register

2.2 Bidirectional Ports

With the exception of Port 6, all ports (0, 1, 4, 5, 7, A and B) are all 4-bit wide. Port 6 has a data width of 2 bits (bit 0 and bit 1). All these ports may be used for data input or output. All ports are equipped with Schmitt-trigger inputs and a variety of mask options for open drain, open source

and full complementary outputs and pull-up and pull-down transistors. The optional pull-up/pull-down transistors are only active when the port is in input mode. All Port Data Registers (PxDAT) are I/O mapped to the primary address register of the respective port address and the Port Data Direction Register (PxDDR), to the corresponding auxiliary register.

Port Data Register (PxDAT)

Primary register address: 'Port address'hex

	Bit 3*	Bit 2	Bit 1	Bit 0	
PxDAT	PxDAT3	PxDAT2	PxDAT1	PxDAT0	Reset value: 1111b

^{*} Bit $3 \rightarrow MSB$, bit $0 \rightarrow LSB$

Port Data Direction Register (PxDDR)

Auxiliary register address: 'Port address'hex

	Bit 3	Bit 2	Bit 1	Bit 0	
PxDDR	PxDDR3	PxDDR2	PxDDR1	PxDDR0	Reset value: 1111b

Value: 1111b means all pins in input mode

Table 5. Port Data Direction Register (PxDDR)

Code	Function
3 2 1 0	
x x x 1	BPx0 in input mode
x x x 0	BPx0 in output mode
x x 1 x	BPx1 in input mode
x x 0 x	BPx1 in output mode
x 1 x x	BPx2 in input mode
x 0 x x	BPx2 in output mode
1 x x x	BPx3 in input mode
0 x x x	BPx3 in output mode

There are five different types of bidirectional ports:

- Type 1 (Ports 0 and 1) 4-bit wide bidirectional ports with automatic full bus width direction switching.
- Type 2 (Ports 5 and 7) 4-bit wide bitwise programmable high drive I/O ports.
- Type 3 (Ports A and B) 4-bit wide bitwise programmable bidirectional ports with optional keyboard pull-ups.
- Type 4 (Port 6) 2-bit wide bitwise programmable bidirectional ports with optional bus pullups and programmable interrupt logic.
- Type 5 (Port 4) 4-bit wide bitwise programmable bidirectional port also provides the I/O interface to Timer 0 and Buzzer.

2.2.1 Port 0, Port 1 – Bidirectional Ports Type 1

In this port type, the data direction register is not independently software programmable, the direction of the complete port being switched automatically when an I/O instruction occurs (see figure 13). The port is switched to output mode with an OUT instruction and to input with an IN instruction. The data written to a port will be stored into the output data latches and appears immediately at the port pin following the OUT instruction. After RESET all output latches are set to '1' and the ports are switched to input mode. An IN instruction reads the condition of the associated pins.

Note

Care must be taken when switching these bidirectional ports from output to input. The capacitive pin loading at this port in conjunction with the high resistance pull-ups may cause the CPU to read the contents of the output data register rather than the external input state. To avoid this, one should use either of the following programming techniques:

- Use two IN instructions and DROP the first data nibble. The first IN switches the port from output to input and the DROP removes the first invalid nibble. The second IN reads the valid pin state.
- Use an OUT instruction followed by an IN instruction.
 With the OUT instruction the capacitive load is charged or discharged depending on the optional pull-up /pull-down configuration. Write a "1" for pins with pull-up resistors and a "0" for pins with pull-down resistors.

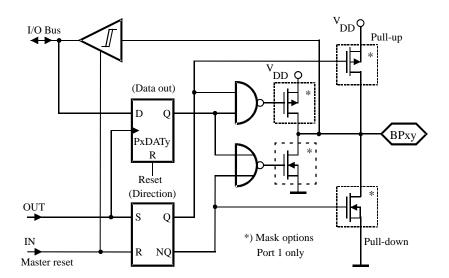


Figure 13. Bidirectional Ports 0 and 1

2.2.2 Port 5, Port 7 – Bidirectional Ports Type 2 Port A, Port B – Bidirectional Ports Type 3

These, and all other bidirectional ports include a bitwise programmable Data Direction Register (PxDDR), which allows the individual programming of each port bit as input or output. It also opens up the possibility of reading the pin condition when in output mode. This is a useful feature for self testing and for serial bus applications.

Both type 2 and type 3 bidirectional ports have the same I/O logic. Type 2 however, has an increased drive capability and type 3, an additional low resistance pull-up as costumer mask option.

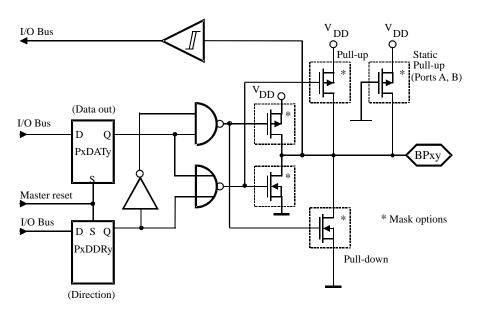


Figure 14. Bidirectional Ports 5, 7, A and B

2.2.3 Port 6 – Bidirectional Port Type 4

This 2-bit bidirectional port can be used as bitwise programmable I/O, the data being LSB aligned so that the two MSB's will not appear on the port pins when written. The port pins can also be used as external interrupt inputs (see figures 15 and 16). Both interrupts can be masked or independently configured to trigger on either edge. The interrupt priority levels are also configurable. The interrupt configuration and port direction is controlled by the Port 6 Control Register (P6CR). An additional low resistance pull-up transistor (mask option) provides an internal bus pull-up for serial bus applications.

In output mode (PxDDR bit = 0), the respective Port Data Register (PxDAT) bit will appear on the port pin, driven by an output port driver stage which can be mask programmed as open drain, or full complementary CMOS. With an IN instruction the actual pin state can at any time be read back into the controller without changing the port directional mode. So, for example should the output port

be mask configured as an open drain driver, as long as the output transistor is off, the controller is able to receive external data on this pin without switching into input mode.

In input mode (PxDDR bit = 1), the output driver stage is deactivated, so that an IN instruction will directly read the pin state which can be driven from an external source. In this case the state of the Port Data Register (PxDAT), although not appearing at the pin itself remains unchanged. High resistance mask selectable pull-up or pull-down transistors are automatically switched onto the port pin in input mode. The Port Data Register is written with an OUT instruction to the respective port address.

The Port 6 Data Register (P6DAT) is I/O mapped to the primary address register of address '6'hex and the Port 6 Control Register (P6CR) to the corresponding auxiliary register. The P6CR is a byte wide register and is written by writing first the low nibble then the high nibble (see section 2.1 "Addressing peripherals").

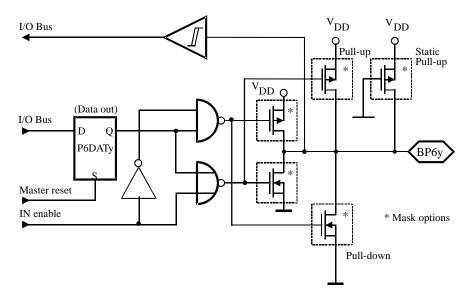


Figure 15. Bidirectional Port 6

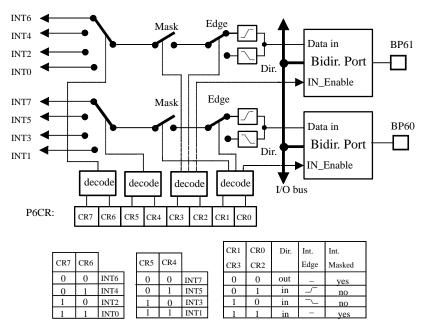


Figure 16. Port 6 external interrupts

Port 6 Data Register (P6DAT)

Primary register address: '6'hex

P6DAT	not used	not used	P6DAT1	P6DAT0	Reset value: xx11b
	B11 3	B 10 2	Bit I	Bit U	

The unused bits 2 and 3 are '0', if read.

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Port 6 Control Register (P6CR)

Auxiliary register address: '6'hex

	Second write cycle	P60PR2	P60PR1	P61PR2	P61PR1	Reset value: 1111b
	G 1 4 1	DC0DD4	DCODD1	DC1DD4	DC1DD1	D 4 1 11111
		Bit 7	Bit 6	Bit 5	Bit 4	
P6CR	First write cycle	P60IM2	P60IM1	P61IM2	P6IM1	Reset value: 1111b
		Bit 3	Bit 2	Bit 1	Bit 0	

P6xIM2, P6xIM1 – Port 6x Interrupt mode/direction code

P6xPR2, P6xPR1 – BP6x Interrupt priority code

Table 6. Port 6 control register (P6CR)

A	auxiliary Address: '6'hex First write cycle		Second write cycle
Code	Function	Code	Function
3 2 1 0		3 2 1 0	
x x 1 1	BP60 in input mode – interrupt masked	x x 1 1	BP60 set to priority 1
x x 0 1	BP60 in input mode – rising edge interrupt	x x 0 1	BP60 set to priority 3
x x 1 0	BP60 in input mode – falling edge interrupt	x x 1 0	BP60 set to priority 5
x x 0 0	BP60 in output mode – interrupt masked	x x 0 0	BP60 set to priority 7
11xx	BP61 in input mode – interrupt masked	11xx	BP61 set to priority 0
0 1 x x	BP61 in input mode – rising edge interrupt	0 1 x x	BP61 set to priority 2
10 x x	BP61 in input mode – falling edge interrupt	10 x x	BP61 set to priority 4
0 0 x x	BP61 in output mode – interrupt masked	0 0 x x	BP61 set to priority 6

2.2.4 Port 4 – Bidirectional Port Type 5

The type 5 bidirectional port is both a bitwise configurable I/O port and provides the external pins for both the Timer 0 and the internal buzzer generator. As a normal port, it performs in exactly the same way as bidirectional

port type 2 (see figure 14). Two additional multiplexers allow data and port direction control to be passed over to other internal modules (Timer 0 or Buzzer). Each of the four Port 4 pins can be individually switched by the Timer/Counter I/O Register (TCIO). Figure 17 shows the internal interfaces to Port 4.

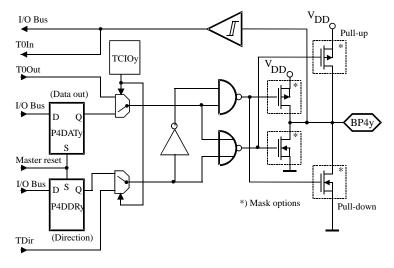


Figure 17. Bidirectional Port 4

2.2.5 TIM1 – Bidirectional Pin Timer 1

TIM1 is a dedicated bidirectional I/O stage for signal communication to and from the Timer 1 in the timer/

counter module (see figure 18). It has no I/O bus interface and is not directly accessible from the CPU. The direction control is performed from the timer/counter configuration registers.

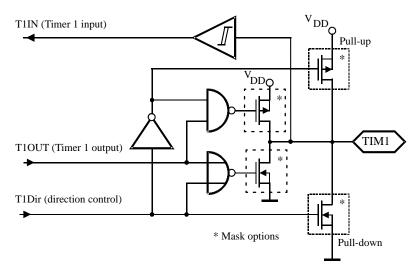


Figure 18. Bidirectional pin TIM1

2.2.6 Port Monitor Module

In addition to the standard I/O functions, both Port A (BPA3 – BPA0) and Port B (BPB3 – BPB0) are equipped with port monitor modules. This module is connected across all four port pins (see figure 19) and generates an interrupt should a preprogrammed transition occur on any of the selected pins. This allows interrupt driven port scanning without the power consuming task of continuously polling the port inputs.

Using the Port Interrupt Control Register (PxICR), pins can be individually selected. A non-selected pin cannot

generate an interrupt. The Port Interrupt Priority Register (PxIPR) allows masking of each interrupt, definition of the interrupt edge and programming of the interrupt priority levels.

The Port Interrupt Priority Registers PAIPR and PBIPR are I/O mapped to the the primary address registers of the Port Monitor Module addresses '2'h and '3'h respectively. The Port Interrupt Control Registers PAICR and PBICR are mapped to the corresponding auxiliary registers

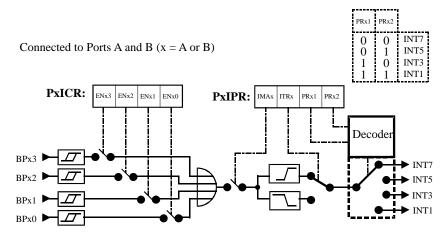


Figure 19. Port monitor module

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Port Monitor Interrupt Priority Register (PxIPR)

x = A' (Port A) or B' (Port B)

(Port A) Primary register address: '2'hex

(Port B) Primary register address: '3'hex

 Bit 3
 Bit 2
 Bit 1
 Bit 0

 PxIPR
 IMx
 ITRx
 PRx2
 PRx1
 Reset value: 1111b

IMx - Interrupt Mask

ITRx - Interrupt Transition

PRx2..1 – Interrupt Priority code

Table 7. Port Monitor Interrupt Priority Register (PxIPR)

Code 3 2 1 0	Function				
x x 0 0	Port monitor interrupt priority 7				
x x 0 1	Port monitor interrupt priority 5				
x x 1 0	Port monitor interrupt priority 3				
x x 1 1	Port monitor interrupt priority 1				
x 0 x x	Port monitor interrupt on rising edge				
x 1 x x	Port monitor interrupt on falling edge				
0 x x x	Port monitor interrupt enabled				
1 x x x	Port monitor interrupt masked				

Port Monitor Interrupt Control Register (PxICR)

x = A' (Port A) or 'B' (Port B)

(Port A) Auxiliary register address: '2'hex

(Port B) Auxiliary register address: '3'hex

 Bit 3
 Bit 2
 Bit 1
 Bit 0

 PxICR
 ENx3
 ENx2
 ENx1
 ENx0
 Reset value: 1111b

ENx3 ... 0 port monitor input ENable code

Table 8. Port Monitor Interrupt Control Register (PxICR)

Code	Function
3 2 1 0	
x x x 0	Bit 0 can generate an interrupt
x x x 1	Bit 0 cannot generate an interrupt
x x 0 x	Bit 1 can generate an interrupt
x x 1 x	Bit 1 cannot generate an interrupt
x 0 x x	Bit 2 can generate an interrupt
x 1 x x	Bit 2 cannot generate an interrupt
0 x x x	Bit 3 can generate an interrupt
1 x x x	Bit 3 cannot generate an interrupt

2.3 Watch Timer

The watch timer is a frequency divider for generating two independent time base interrupts. It is driven by the SUBCL generated by the clock module (see figure 10) and consists of a 15 stage binary divider and two programmable multiplexers for selecting the appropriate interrupt frequencies for each interrupt source (see figure 20). Each multiplexer is completely independent and is controlled

by the common Watch Timer Frequency Select Register (WTFSR). Buffer registers store the respective frequency select codes and ensure complete programming independence of each interrupt channel.

Interrupt masking and programming of the interrupt priority levels is performed with the aid of the Watch Timer Interrupt Priority Register (WTIPR).

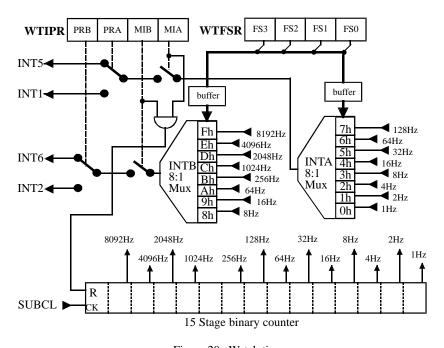


Figure 20. Watch timer

2.3.1 Watch Timer Registers

The Watch Timer Frequency Select Register (WTFSR) is I/O mapped to the primary address register of the watch timer address ('F'hex) and the Watch Timer Interrupt Priority Register (WTIPR) to the corresponding auxiliary register.

FS3 determines whether the INTA or the INTB buffer register is loaded with the select code (FS2-FS0).

The interrupt masks MIA and MIB enable interrupt masking of INTA and INTB respectively. Each interrupt source can be programmed with PRA and PRB to one of two interrupt priority levels. Disabling both interrupts resets the watch timer.

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Watch Timer Interrupt Priority Register (WTIPR)

Auxiliary register address (write only): 'F'hex

WTIPR	PRR	PRA	MIB	MIA	Reset value: 1111b
** ***	LVD	IIA	MIID	IVIIA	Reset value. 11110

PRB - Watch Timer Interrupt INTB Priority select

PRA – Watch Timer Interrupt INTB Priority select

MIB – Watch Timer Interrupt INTB Mask

MIA – Watch Timer Interrupt INTA Mask

Table 9. Watch Timer Interrupt Priority Register (WTIPR)

Code	Function
3 2 1 0	
x x x 1	Mask interrupt A
x x x 0	Enable interrupt A
x x 1 x	Mask interrupt B
x x 0 x	Enable interrupt B
x 1 x x	Interrupt A => priority 1
x 0 x x	Interrupt A => priority 5
1 x x x	Interrupt B => priority 2
0 x x x	Interrupt B => priority 6
x x 1 1	Reset prescaler and halt

Watch Timer Frequency Select Register (WTFSR)

Primary register address (write only): 'F'hex

WTFSR	FS3	FS2	FS1	FS0	Reset value: 1111b
	Bit 3	Bit 2	Bit I	Bit 0	

FS3 ... 0 – Frequency Select code

Table 10. Watch Timer Frequency Select Register (WTFSR)

Code 3 2 1 0	Function		Code 3 2 1 0	Function	
0000	INTA	Select 1 Hz	1000	INTB	Select 8 Hz
0001		Select 2 Hz	1001		Select 16 Hz
0010		Select 4 Hz	1010		Select 64 Hz
0011		Select 8 Hz	1011		Select 256 Hz
0100		Select 16 Hz	1100		Select 1024 Hz
0101		Select 32 Hz	1101		Select 2048 Hz
0110		Select 64 Hz	1110		Select 4096 Hz
0111		Select 128 Hz	1111		Select 8192 Hz

2.4 Watchdog Timer

The Watchdog timer is a 17 stage binary divider clocked by SUBCL generated within the clock module (see figures 10 and 21). It can only be enabled as a mask option whereby it must be periodically reset from the application program. The program cannot disable the watchdog. Should for any reason, the CPU find itself for an extended period in SLEEP or in a section of program that includes

no watchdog reset, then the watchdog will overflow, thus forcing the NRST pin low. This, in turn initiates a master reset. The timeout period can be set to 0.5, 1 or 2 seconds as a mask option.

To reset the watchdog the program must perform an IN-instruction on the address WDRES ('3'hex). No relevant data will be received, so this operation will normally followed by a DROP to flush the data from the stack.

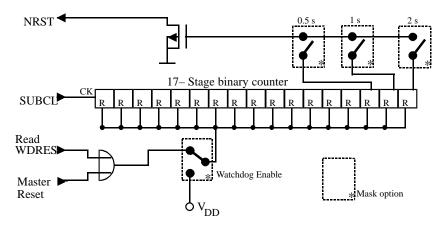


Figure 21. Watchdog timer

2.5 Timer/Counter Module (TCM)

The TCM consists of two timer/counter blocks (Timer 0 and Timer 1), which can be used separately, or together as a single 16-bit counter/timer (see figure 22). Each timer can be supplied by various internal or external clock sources, which can be selected and predivided under program control using the Timer/Counter Control Register (TCCR), the Timer 0 Control Register (T0CR) and the Timer 1 Control Register (T1CR). Capture and compare registers (T0CA,T1CA,T0CP and T1CP) allow not only event counting but also the generation of various timed output waveforms including programmable frequencies, modulated melody tones, Pulse Width Modulated (PWM) and Pulse Density Modulated (PDM) output signals. When in one of these signal generation modes, the capture registers act as timer shadow registers, freezing the current timer state whenever read by the CPU. The Timer 0

is further equipped for performing a variety of time measurement operations, using the capture register together with the gating logic for performing asynchronous, externally triggered snapshot measurements. These measurements include single input pulse width and period measurements and also dual input phase and positional measurement. The mode configuration is set in the Timer 0 and Timer 1 Mode Registers (TOMO and T1MO).

Each timer represents a single maskable interrupt source (T0INT and T1INT), the priority of which can be configured under program control. A Timer 0 interrupt can be caused by any of three conditions (overflow, compare or end-of-measurement). The associated status register (T0SR) differentiates between these. No status register is necessary in the Timer 1 as an interrupt is caused only on a compare condition.

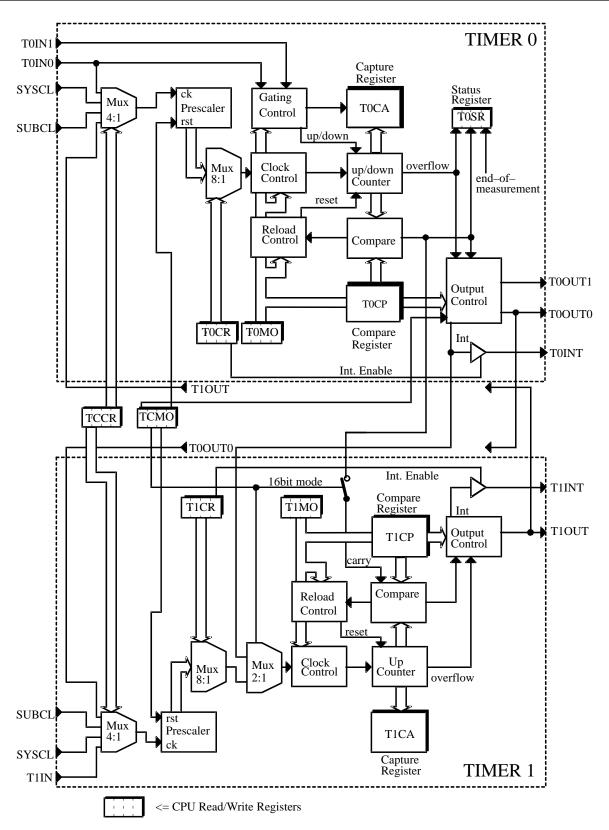


Figure 22. Timer/counter module

2.5.1 **Timer/Counter Registers**

With the exception of the Timer 0 Interrupt Status Register (TOSR), all the timer/counter registers are indirectly addressed using extended addressing as described in section "Addressing peripherals". An overview of all register and subport addresses is shown in table 4. The Timer/ Counter auxiliary register (TCX) holds the subport address of the particular register about to be accessed.

Timer 0 Interrupt Status Register (T0SR)

Auxiliary register address (read access): '9'hex

T0SR	not used	T0EOM	T0OFL	T0CMP	Reset value: x000b
	Bit 3	Bit 2	Bit 1	Bit 0	

Note: The status register is reset automatically when read and also when the Timer 0 is reset.

T0EOM - Timer 0 End Of Measurement status flag

TOOFL - Timer 0 Over FLow status flag

T0CMP – Timer 0 CoMPare status flag

Table 11. Timer 0 Interrupt Status Register (T0SR)

Code 3 2 1 0	Function			
x x x 1	Timer 0 compare has occurred (Timer $0 = T0CP$)			
x x 1 x	Timer 0 overflow or underflow has occurred			
x 1 x x	Timer 0 measurement completed			

The interrupt flags will be set whenever the associated condition occurs irrespective of whether the corresponding interrupt is triggered. So, the status flags will still be set if the interrupt condition occurs when the interrupt is masked. To see exactly when the flags are set, see T0MO control code table.

Reading from the timer/counter auxiliary register will access the Timer 0 Interrupt Status Register (T0SR).

Timer 0 Compare Register (T0CP) – Byte Write

Subport address (write access): '9'hex

		Bit 3	Bit 2	Bit 1	Bit 0	_
T0CP	First write cycle	T0CP3	T0CP2	T0CP1	T0CP0	Reset value: xxxxb
		Bit 7	Bit 6	Bit 5	Bit 4	_
	Second write cycle	T0CP7	T0CP6	T0CP5	T0CP4	Reset value: xxxxb

T0CP3 ... T0CP0 - Timer 0 Compare Register Data (low nibble) - first write cycle

T0CP7 ... T0CP4 - Timer 0 Compare Register Data (high nibble) - second write cycle

Timer/Counter Compare Register (T0CP)					
Subport address: '9'hex	(Indirect write access)				
Timer/Counter Subport Pointer (TCX) Address: '9'hex					

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Timer 1 Compare Register (T1CP) - Byte Write

Subport address (write access): '8'hex

Bit 0 Bit 3 Bit 2 Bit 1 T1CP First write cycle T1CP2 T1CP1 T1CP3 T1CP0 Bit 7 Bit 6 Bit 5 Bit 4

Reset value: xxxxb

Second write cycle

T1CP7 T1CP6 T1CP5 T1CP4

Reset value: xxxxb

T1CP3 ... T1CP0 - Timer 1 Compare Register Data (low nibble) - first write cycle

T1CP7. .. T1CP4 – Timer 1 Compare Register Data (high nibble) – second write cycle

Timer/Counter Compare Register (T1CP)					
Subport address: '8'hex	(indirect write access)				
Timer/Counter Subport Po	inter (TCX) Address: '9'hex				

The compare registers (TOCP and T1CP) are both 8-bits wide and must be accessed as byte wide subports (see section "Addressing Peripherals"). They are written low nibble first followed by the high nibble. Any timer interrupts are automatically suppressed until the complete compare value has been transferred.

Timer 0 Capture Register (T0CA) - Byte Read

Subport address (read access): '9'hex

Bit 7 Bit 5 Bit 4 Bit 6 T₀CA First read cycle T0CA7 T0CA6 T0CA5 T0CA4 Reset value: 0000b Bit 3 Bit 2 Bit 1 Bit 0 Second read cycle T0CA3 T0CA2 T0CA1 T₀CA₀ Reset value: 0000b

T0CA7. .. T0CA4 – Timer 0 Capture Register Data (high nibble) – first read cycle

T0CA3 ... T0CA0 - Timer 0 Capture Register Data (low nibble) - second read cycle

Note: In PDM mode only, if the timer is read, the bit order will appear reversed, so that T0CA0 = MSB, $TOCA1=MSB-1 \dots TOCA6=LSB+1, TOCA7=LSB.$

Timer/Counter Capture Register (T0CA)					
Subport address: '9'hex	(indirect read access)				
Timer/Counter Subport Pointer (TCX) Address: '9'hex					



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Timer 1 Capture Register (T1CA) - Byte Read

Subport address (read access): '8'hex

		Bit 7	Bit 6	Bit 5	Bit 4	
T1CA	First read cycle	T1CA7	T1CA6	T1CA5	T1CA4	Reset value: 0000b
		Bit 3	Bit 2	Bit 1	Bit 0	
	Second read cycle	T1CA3	T1CA2	T1CA1	T1CA0	Reset value: 0000b

T1CA7 ... T1CA4 – Timer 1 Capture Register Data (high nibble) – first read cycle

T1CA3 ... T1CA0 – Timer 1 Capture Register Data (low nibble) – second read cycle

Timer/Counter Capture Register (T1CA)					
Subport address: '8'hex	(indirect read access)				
Timer/Counter Subport Pointer (TCX) Address: '9'hex					

The 8-bit capture registers (T0CA and T1CA) are read as byte wide subports. Note however, unlike the writing to the compare registers, the high nibble is read first followed by the low nibble. The 8-bit timer state is captured

on reading the first nibble and held until the complete byte has been read. During this transfer the timer is free to continue counting.

Timer/Counter Control Register (TCCR)

Subport address (indirect access): '6'hex

TO COD	Bit 3	Bit 2	Bit I	Bit 0	
TCCR	T0CL2	T0CL1	T1CL2	T1CL1	Reset value: 1111b

T0CL2, T0CL1 - Timer 0 prescaler clock source select code

T1CL2, T1CL1 - Timer 1 prescaler clock source select code

Table 12. Timer/Counter Control Register (TCCR)

	Timer/Counter Subport Pointer (TCX) Address: 9h						
Code	Function	Directio	on (TDir)				
3 2 1 0		BP40*	TIM1				
x x 0 0	Timer 0 prescaler clock =SUBCL (32 kHz)	out	X				
x x 0 1	Timer 0 prescaler clock =SYSCL (2 MHz)	out	X				
x x 1 0	Timer 0 prescaler clock = Timer1 output (T1OUT)	out	X				
x x 1 1	Timer 0 prescaler clock =T0IN0 (BP40*)	in	x				
0 0 x x	Timer 1 prescaler clock =SUBCL (32 kHz)	X	out				
0 1 x x	Timer 1 prescaler clock =SYSCL (2 MHz)	X	out				
10 x x	Timer 1 prescaler clock = Timer 0 output (T0OUT0)	X	out				
1 1 x x	Timer 1 prescaler clock =TIM1	X	in				

^{*} if TCIO0 = low (connects Timer 0 to Port 4)

The Timer/Counter Control Register (TCCR) controls the clock source to both Timer 0 and Timer 1 prescalers. If an external clock source (on BP40 or TIM1) is selected, then

the corresponding port direction is automatically switched to input mode (see figure 23).

Note: The TCIO0 bit must be set low for the BP40 external timer/counter access.

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Timer/Counter I/O Control Register (TCIOR)

D:/ 2

Subport address (indirect access): '5'hex

	ב וום	DIL Z	DIL I	סונט	
TCIOR	TCIO3	TCIO2	TCIO1	TCIO0	Reset value: 1111b

D:40

D'4 1

Table 13. Timer/Counter I/O Control Register (TCIOR)

	Timer/Counter Subport Pointer (TCX) Address: '9'hex				
Code	Function				
3 2 1 0					
x x x 1	BP40 – standard port mode				
x x x 0	BP40 – Timer 0 clock input (T0IN0) or Timer 0 output (T0OUT0)				
x x 1 x	BP41 – standard port mode				
x x 0 x	BP41 – Timer 0 gate input (T0IN1) or Timer 0 output (T0OUT1)				
x 1 x x	BP42 – standard port mode				
x 0 x x	BP42 – Buzzer output (BUZ)				
1 x x x	BP43 – standard port mode				
0 x x x	BP43 – Buzzer output (NBUZ)				

Using the Timer/Counter I/O Control Register (TCIOR) the program can configure the respective Port 4 pins as either standard data I/O ports or as external signal ports for the Timer 0 and Buzzer. The Timer 1 uses a dedicated I/O pin TIM1, the direction of which is controlled solely by the TCCR (see figure 23). It should be noticed that if a

TCIOR bit is set low, then the corresponding port data direction register (P4DDR) bit no longer influences the port direction. In the case of BP40 and BP41, the port direction is then controlled entirely by the timer/counter configuration registers (TCCR,T0MO), while pins BP42 and BP43 become unidirectional buzzer outputs.

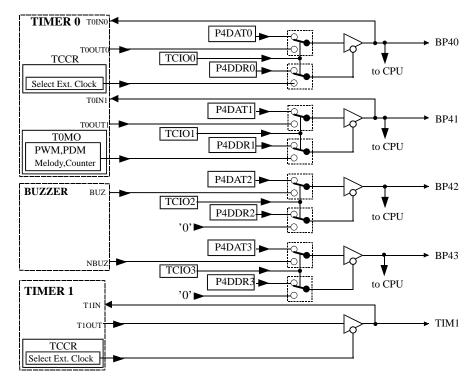


Figure 23. Timer/counter and buzzer external interface

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Timer 0 Control Register (T0CR)

The TOCR is responsible for the predivision of the selected Timer 0 input clock. It can be divided or used di-

rectly as clock for the up/down counter. Bit 0 is the mask bit for the Timer 0 interrupt.

Subport address (indirect access): '1'hex

	Bit 3	Bit 2	Bit I	Bit 0	
T0CR	T0FS3	T0FS2	T0FS1	T0IM	Reset value: 1111b

T0FS3 ... 1 – Timer 0 prescaler division factor code

T0IM - Timer 0 Interrupt Mask

Table 14. Timer 0 Control Register (T0CR)

	Timer/Counter Subport Pointer (TCX) Address: '9'hex				
Code	Function				
3 2 1 0					
x x x 1	Timer 0 interrupt masked				
x x x 0	Timer 0 interrupt enabled				
0 0 0 x	Timer 0 prescaler divide by 256				
0 0 1 x	Timer 0 prescaler divide by 128				
0 1 0 x	Timer 0 prescaler divide by 64				
0 1 1 x	Timer 0 prescaler divide by 32				
1 0 0 x	Timer 0 prescaler divide by 16				
1 0 1 x	Timer 0 prescaler divide by 8				
1 1 0 x	Timer 0 prescaler divide by 2				
1 1 1 x	Timer 0 prescaler bypassed				

M44C510

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Timer 1 Control Register (T1CR)

The T1CR is responsible for the predivision of the selected Timer 1 input clock. It can be divided or used

directly as clock for the up counter. Bit 0 is the mask bit for the Timer 1 interrupt.

Subport address (indirect access): '3'hex

 Bit 3
 Bit 2
 Bit 1
 Bit 0

 T1CR
 T1FS3
 T1FS2
 T1FS1
 T1IM
 Reset value: 1111b

T0FS3 ... 1 - Timer 1 prescaler division factor code

T0IM - Timer 1 Interrupt Mask

Table 15. Timer 1 Control Register (T1CR)

Timer/Counter Subport Pointer (TCX) Address: '9'hex				
Code	Function			
3 2 1 0				
x x x 1	Timer 1 interrupt masked			
x x x 0	Timer 1 interrupt enabled			
0 0 0 x	Timer 1 prescaler divide by 256			
0 0 1 x	Timer 1 prescaler divide by 128			
0 1 0 x	Timer 1 prescaler divide by 64			
0 1 1 x	Timer 1 prescaler divide by 32			
1 0 0 x	Timer 1 prescaler divide by 16			
1 0 1 x	Timer 1 prescaler divide by 8			
1 1 0 x	Timer 1 prescaler divide by 2			
111x	Timer 1 prescaler bypassed			

Timer/Counter Interrupt Priority Register (TCIP)

The Timer/Counter Interrupt Priority register (TCIP) is used to configure the Timer 0 and Timer 1 interrupt priority levels.

Subport address (indirect access): '7'hex

 Bit 3
 Bit 2
 Bit 1
 Bit 0

 TCIP
 T1IP2
 T1IP11
 T0IP2
 T0IP1
 Reset value: 1111b

T0IP2, T0IP1 - Timer 0 Interrupt Priority code

T1IP2, T1IP1 - Timer 1 Interrupt Priority code

Table 16. Timer/Counter Interrupt Priority Register (TCIP)

	Timer/Counter Subport Pointer (TCX) Address: '9'hex					
Code	Function					
3 2 1 0						
x x 1 1	Timer 0 interrupt priority 1					
x x 1 0	Timer 0 interrupt priority 3					
x x 0 1	Timer 0 interrupt priority 5					
x x 0 0	Timer 0 interrupt priority 7					
1 1 x x	Timer 1 interrupt priority 0					
1 0 x x	Timer 1 interrupt priority 2					
0 1 x x	Timer 1 interrupt priority 4					
0 0 x x	Timer 1 interrupt priority 6					

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2.5.2 Timer 0 Modes

The Timer 0 mode configuration is defined in the Timer 0 Mode Register (T0MO). The available modes and the effect on the Timer 0 interrupt and interrupt flags is shown below. In all modes except the position measurement mode, the Timer 0 performs as an up-counter, the related clock frequency being defined by the selected clock source and the prescaler division factor. The counter can be reset and halted at any time by the T0RST bit of the TCMO register which, in turn also resets all the interrupt status flags and capture registers. Whenever the Port 4

BP40 and BP41 pins are required for Timer 0 I/O then the appropriate TIOR enable bit must be set low. In which case the port direction switching is handled automatically by the hardware. In modes where the BP40 is not used as a timer clock input or as a melody envelope output, it outputs the same signal as that appearing on BP41. With the help of the T0NINV bit of the Timer/Counter Mode Register (TCMO), the BP41 output can be inverted so that BP40 and BP41 form a differential output stage which can be used for direct driving of piezo buzzers or small stepper motors.

Timer 0 Mode Register (T0MO)

Subport address (indirect access): '0'hex

T0MO3 ... 0 – Timer 0 Mode Code

Table 17. Timer 0 Mode Register (T0MO)

Timer/Counter Subport Pointer (TCX) Address: '9'hex							
Code	Function	Assuming TCIOR1=TCIOR0=low		Interrupts(Flag)			
3 2 1 0		77.10	DD44 (#2)	-			
		BP40	BP41(*3)	cmp	ofl	eom	
0000	reserved			_	-	_	
0001	reserved			_	-	_	
0010	Modulated melody mode	Envelope(out)	Tone(out)	y(y)	y(y)	n(n)	
0011	Melody mode	Tone(out)	Tone(out)	y(y)	y(y)	n(n)	
0100	Counter-auto reload (50% duty cycle)	Toggle(out)/Clock(in)	Toggle(out)	y(y)	y(y)	n(n)	
0101	Counter-free running (50% duty cycle)	Toggle(out)/Clock(in)	Toggle(out)	n(y)	y(y)	n(n)	
0110	Pulse density modulation	PDM(out)/Clock(in)	PDM(out)	n(y)	y(y)	n(n)	
0111	Pulse width modulation	PWM(out)/Clock(in)	PWM(out)	n(y)	y(y)	n(n)	
1000	Phase measurement	Signal 1(in)	Signal 2(in)	n(n)	y(y)	y(y)	
1001	Position measurement	Signal 1(in)	Signal 2(in)	(*1)	(*2)	n(n)	
1010	Low pulse width measurement	Clock(in)	Signal(in)	n(y)	y(y)	y(y)	
1011	High pulse width measurement	Clock(in)	Signal(in)	n(y)	y(y)	y(y)	
1100	Counter- auto reload (strobe)	Strobe(out)/Clock(in)	Strobe(out)	y(y)	y(y)	n(y)	
1101	Counter-free running (strobe)	Strobe(out)/Clock(in)	Strobe(out)	n(y)	y(y)	n(y)	
1110	Period measurement (rising edge)	Clock(in)	Signal(in)	n(y)	y(y)	y(y)	
1111	Period measurement (falling edge)	Clock(in)	Signal(in)	n(y)	y(y)	y(y)	

*1 **Note:** The compare interrupt/status flag can only be set when counting up.

*2 **Note:** The overflow interrupt/status flag is set on both an overflow or an underflow.

*3 **Note:** The BP41 signals can be inverted if T0NINV=0 (TCMO register)

Timer 0 Free Running Counter Modes (Strobe and 50% Duty Cycle)

In the free running counter mode, the Timer 0 can be used as an event counter for summing external event pulses on BP40 or as a timer with an internal time base clock. When enabled, the counter will count up, generating an output signal on BP41 whenever the counter contents match the compare register (see figure 24). This signal can, depending on the timer mode, appear either as a strobe pulse or

as a simple toggling of the output state (50% duty cycle). Interrupts, (if not masked) are generated every 256 clocks on the overflow condition. By reading the capture register, the current counter state can be read at any time. The compare register has no effect on the counter cycle time and will not influence interrupts.

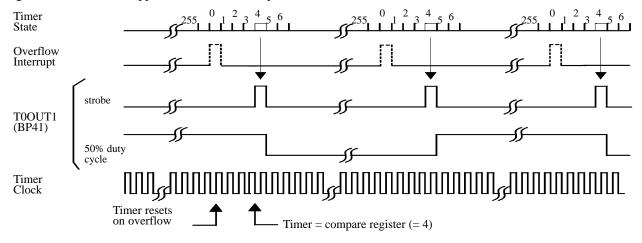


Figure 24. Timer 0 free running counter mode

Timer 0 Counter Reload Modes (Strobe and 50% Duty Cycle)

As in the free running mode, the counter can also be clocked from either an external signal on BP40 or from an internal clock source. In this mode, the counter repetition period is completely defined by the contents of the compare register (T0CP) (see figure 25). The counter counts up with the selected clock frequency. When it reaches the value held in the compare register, it causes

the counter to return to the zero state. At the same time, depending on the selected timer mode, the BP41 either toggles or generates a strobe pulse. If the Timer 0 interrupt is unmasked, a compare interrupt is also generated.

The resultant output frequency $f_{OUT} = f_{IN}/2*(n+1)$ where n = compare value (n = 1 - 255).

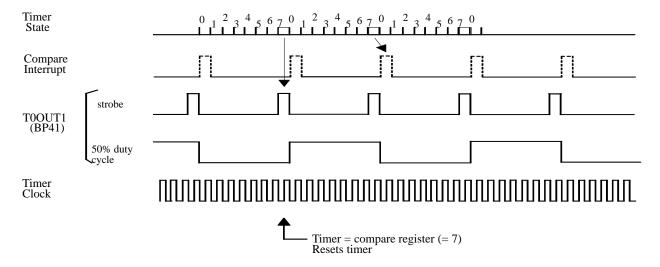


Figure 25. Timer 0 counter reload mode

Motor Chopping and Mask Options

In the counter auto reload mode (50% duty cycle), mask options are available for generating a 1 kHz or 2 kHz frequency with duty cycles of 1/2, 3/8, 5/8 and 3/4. The resultant waveform is used as the chopping frequency for so called "motor chopping". This technique allows the use of low cost, low voltage clock motors in applications where only higher supply voltages are available. The resultant voltage waveforms are shown in figure 26. To obtain the required motor driver waveforms on BP40 and BP41 as shown in figure 27, the user program must modu-

late the Timer 0 chopping frequency. This is performed by preloading Port 4 data latches (P4DAT0 and P4DAT1) with '0', setting the normal Port 4 direction register bits to output mode (P4DDR0 = P4DDR1 = '0') and switching the TCIO0 and TCIO1 register bits alternately to '0', on every chopping burst. The timer chopping signals are thus transferred to the port outputs. In the intermediate periods, between bursts, both TCIO0 and TCIO1 are set to '1' and the preloaded Port 4 data latch outputs appear on the port outputs.

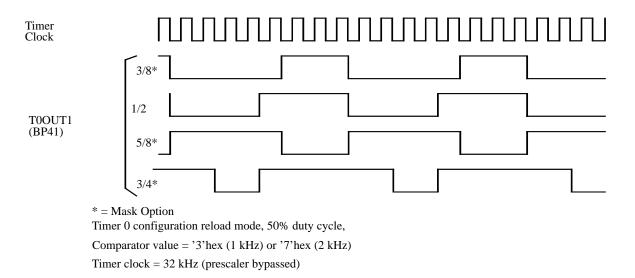


Figure 26. Motor chopping waveforms

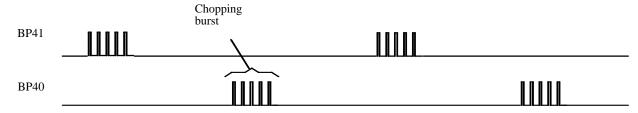


Figure 27. Motor driver output waveforms

Melody Mode (with/without Modulation)

The non modulated melody mode is identical to the auto reload counter (50% duty cycle) mode. The melody tone frequency appearing on BP41 and/or BP40 is determined in exactly the same way by the value written into the comparator register. In the modulated melody mode, the M44C510 generates two output signals, a melody tone and an envelope pulse (see figure 28). The tone frequency output on BP41 is generated in exactly the same way as

in the simple melody mode. While the envelope pulse on BP40 is a single pulse, of a clock period in duration which appears shortly after loading the compare value into the compare register. In this mode, an analogue switch is activated between the BP40 and BP41 outputs (see figure 29). With the external capacitor connected, the resultant signal on BP41 exhibits a melody chime effect with an exponential decay.

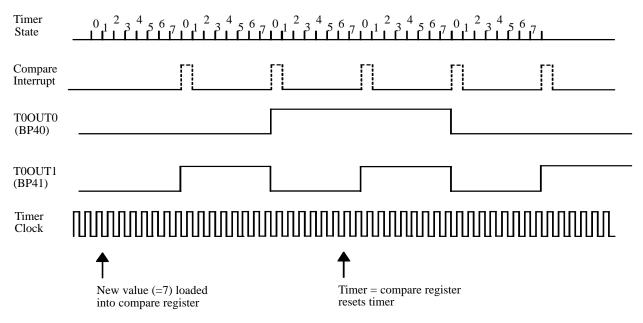


Figure 28. Modulated melody mode

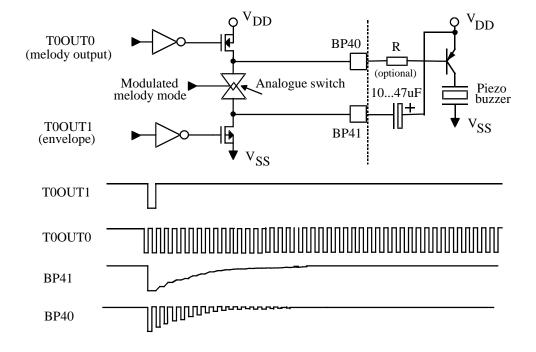


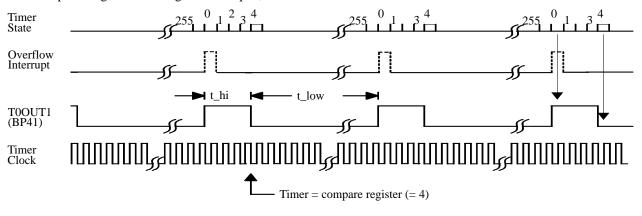
Figure 29. Modulated melody output circuit

Timer 0 Pulse Width Modulation Mode

A pulse width modulated (PWM) signal exhibits a fixed repetition frequency and a variable mark space ratio. It is often used as a simple method for D/A conversion, the high period is proportional to the digital value to be converted. Thus by connecting a simple low pass RC network on the PWM signal, the DC analogue value can be extracted.

The Timer 0 generates the PWM signal by comparing the state of the free running up counter with the contents of the compare register (see figure 30). If it is less, then the BP41 output is high and if it is greater or equal, then it is

set low. Thus, the high phase of the PWM signal is directly proportional to the compare register contents. A total of 256 possible discrete mark space ratios can be generated ranging from a continuous low signal over a variable pulse width signal to a continuous high signal. The PWM signal has a repetition period of 256 clock periods, an interrupt (if unmasked) being generated on every overflow event. Care should be taken if the SYSCL clock is used as the PWM clock source because it will stop if the CPU goes into SLEEP.



t_hi = (comparator value)*clock period t_low = (255-comparator value)*clock period

Figure 30. Timer 0 pulse width modulation

Pulse Density Modulation Mode

Pulse density modulation (PDM) is also used for simple D/A conversion. Unlike the PWM signal, where the high and low signal phases are always continuous during a single repetition cycle, the PDM distributes these evenly as a series of pulses (see figure 31). This has the advantage that, if used together with an RC smoothing filter for D/A conversion, either the ripple is less than the PWM, or, for

a corresponding ripple error, the filter components can be smaller or the clock frequency lower. To generate the PDM output on BP41, the pulse density is controlled by the contents of the compare register in the same way as the PWM generation. Each of the pulses has a width equal to the counter clock period.

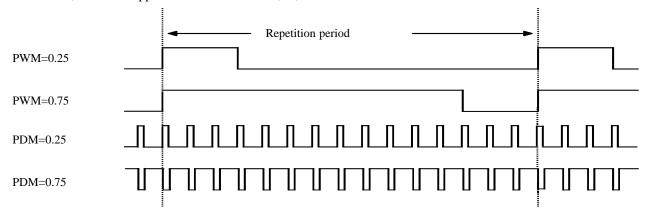


Figure 31. An example 4-bit PWM/PDM comparison

Period Measurement Modes (Rising and Falling Edge)

During the period measurement mode, the counter counts the number of either internal or external clocks in one period of the BP41 input signal (see figure 32). Dependent on the mode chosen, this will be from rising edge to the next rising edge or conversely, falling edge to the following falling edge. On the trigger edge, the counter state will be loaded into the capture register and subsequently reset.

The measurement value remains in the capture register until overwritten by the following measurement value. Interrupts can be generated by either an overflow condition or an end-of-measurement (eom) event. An eom event signals the CPU that a new measurement value is present in the capture register to be read if required.

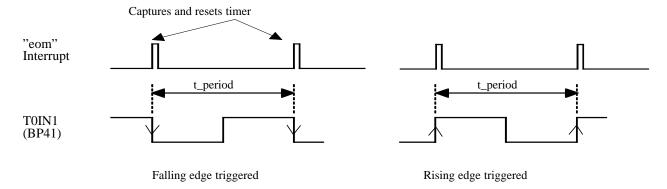


Figure 32. Period measurement

Pulse Width Measurement Modes (High and Low)

In this mode, the selected clock source is gated to the counter for the duration of each input pulse received on BP41 (see figure 33). Whether the measurement takes place during the high or low phase depends on the selected mode. At the end of each pulse, the counter state

is loaded into the capture register and subsequently reset. Interrupts can be generated by either an overflow condition or an end-of-measurement (eom) event. An 'eom' event signals the CPU that a new measurement value is present in the capture register to be read, if required.

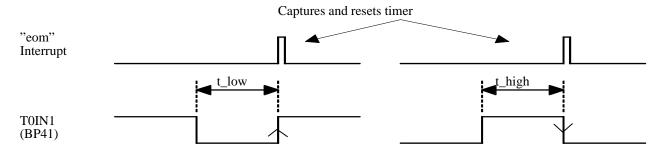


Figure 33. Pulse width measurement

Phase Measurement Mode

This mode allows the Timer 0 to measure the phase misalignment between two 1:1 mark space ratio input signals connected to the BP40 and BP41 pins (see figure 34). The counter clock is gated with the phase misalignment period (tp), during which time the counter increments with the selected clock frequency. This misalignment period is defined as the period during which BP40 is high and BP41 is low. Capturing and resetting of the counter always takes

place on the rising edge of BP41. The measured value remains in the capture register until overwritten by the next measurement. Interrupts can be generated by either an overflow condition or an end-of-measurement ('eom') event. An 'eom' event signals the CPU that a new measurement value is present in the capture register to be read, if required.

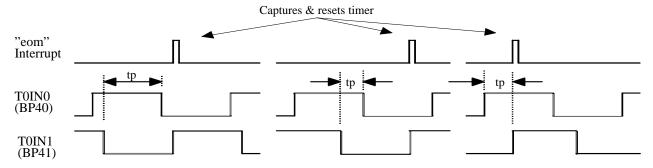


Figure 34. Phase measurement

Position Measurement Mode

This mode is intended for the evaluation of positional sensors with biphase output signals. Figure 35 illustrates a typical positional sensor system which delivers both incremental positional stepping signals and also directional information. The direction can be deduced from the relative phase of the two signals, so that if BP40 is high on the rising edge BP41, then the moving mask is travelling to the left and if it is low then it is travelling right. The direction (left/right) information is used to set the direction of the up/down counter with which the BP40 pulses are counted. So assuming that the system has been reset on a reference position, the counter will always hold the abso-

lute current position of the moving mask. This can be read by the CPU when required. This mode is the only one in which the counter is allowed to decrement. So, in this case it is possible for both an underflow or an overflow to occur. The overflow interrupt (if unmasked) will trigger on either of these conditions while the compare interrupt on the other hand will only trigger if the counter is counting upwards. To differentiate between an overflow or underflow, the compare value can be set to '0'hex for example. An overflow would then set both the overflow and compare status flags while an underflow sets the overflow status flag only.

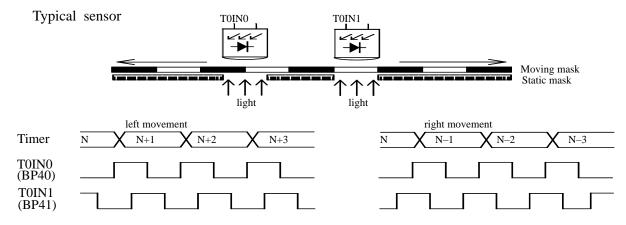


Figure 35. Position measurement mode

2.5.3 Timer 1 Modes

The Timer 1 is aimed at performing event counting and timing functions (see figure 22). It has, unlike the Timer 0 no gated clock or externally triggered capture modes. The counter counts up with either an internal or external clock depending on the state of the Timer 1 Control Register (T1CR) and the Timer/Counter Control Register (TCCR) and generates a compare interrupt whenever the counter matches the Timer 1 compare register. This is the only

Timer 1 interrupt source. Masking can be performed using the mask bit in the Timer 1 Control Register (T1CR) and priority can be defined in the Timer/Counter Interrupt Priority Register (TCIPR). The TIM1 pin is used by the Timer 1 either as clock/event input or timer output. I/O control of the Timer 1 pin TIM1 is controlled entirely by the hardware, so if the TIM1 is selected as an external clock or event source (in the TCCR), then there can be no Timer 1 signal output. In this case, the timer would be used solely to generate interrupts.

Timer 1 Mode Register (T1MO)

Subport address (indirect address): '2'hex

	Bit 3	B1t 2	Bit I	B1t 0	
T1MO	T1MO3	T1MO2	T1MO1	T1MO0	Reset value: 1111b

T0MO3 ... 0 – Timer 1 Mode Code

Table 18. Timer 1 Mode Register (T1MO)

	Timer/Counter Subport Pointer (TCX) Address: '9'hex					
Code	Function	Compare				
3 2 1 0		Interrupt				
1 x 0 0	Counter free running (50% duty cycle)	yes				
1 x 0 1	Counter auto reload (50% duty cycle)	yes				
1 x 1 0	Pulse width modulation	yes				
1 x 1 1	Counter auto-reload (strobe output)	yes				
x 0 x x	Increment on falling edge of clock	_				
x 1 x x	Increment on rising edge of clock	_				
0 x x x	reserved	_				

Timer 1 Counter Free Running (50% Duty Cycle)

In the free running counter mode, the counter counts up with either an internal or external clock and cycles through all 256 timer states. On the clock following a

match between the compare register (T1CR) and the counter, a compare interrupt (if unmasked) is generated and the TIM1 pin is toggled (see figure 36).

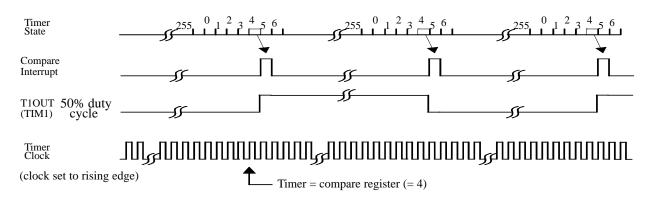


Figure 36. Timer 1 counter free running (50% duty cycle)

Timer 1 Counter Auto Reload (Strobe and 50% Duty Cycle)

In the auto-reload mode, the counter counts up with either an internal or external clock. On the clock cycle following a match between the compare register (T1CR) and the counter, a compare interrupt (if unmasked) is generated, the TIM1 output is either strobed or toggled and the

counter reset (see figure 37). So the counter cycle period is defined by the contents of the compare register. If in 50% duty cycle mode, then the frequency of TIM1

$$\begin{split} f_{TIM1} &= f_{in}/2exp(n{+}1) \\ where the compare value (n) &= 1 \dots 255. \end{split}$$

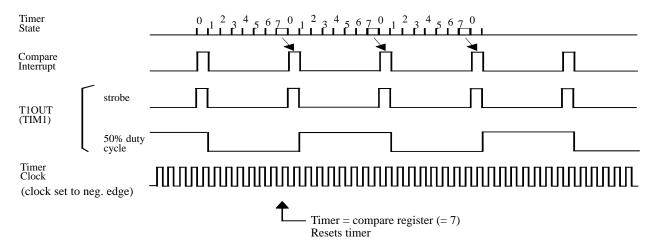
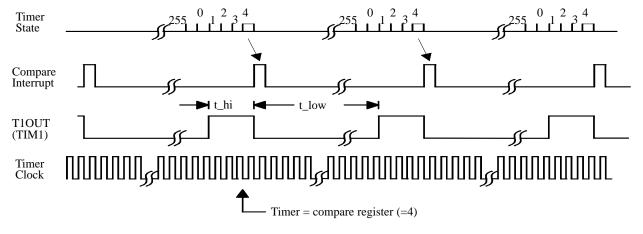


Figure 37. Timer 1 counter auto reload

Timer 1 Pulse Width Modulation

The Timer 1 generates the PWM signal by comparing the state of the free running up counter with the contents of the compare register (see figure 38). If it is less or equal, then the TIM1 output is high and if it is greater, then it is set low. Thus, the high phase of the PWM signal is directly proportional to the compare register contents. A total of 256 possible discrete mark space ratios can be generated

ranging from a continuous low signal over a variable pulse width signal to a continuous high signal. The PWM signal has a repetition period of 256 clock periods, an interrupt (if unmasked) being generated on every compare event. Care should be taken if the SYSCL clock is used as the PWM clock source because it will stop if the CPU goes into SLEEP.



t_hi = (comparator value)*clock period t_low = (255-comparator value)*clock period

Figure 38. Timer 1 pulse width modulation

Timer/Counter Mode Register (TCMO)

Subport address (indirect access): '4'hex

	Bit 3	Bit 2	Bit I	Bit 0	1
TCMO	TONINV	TC8	T1RST	T0RST	Reset value: 1111b

TONINV – Timer 0 Output (BP40) appears non inverted on BP41

TC8 – Timer/Counter 8-/16-bit mode

T1RST - Timer 1 Reset/Run

T0RST - Timer 0 Reset/Run

Table 19. Timer/Counter Mode Register (TCMO)

Timer/Counter Subport Pointer (TCX) Address: '9'hex				
Code	Function			
3 2 1 0				
x x x 0	Timer 0 running			
x x x 1	Timer 0 reset and halted			
x x 0 x	Timer 1 running			
x x 1 x	Timer 1 reset and halted			
x 0 x x	Timer/counter in 16-bit mode			
x 1 x x	Timer/counter in 8-bit mode			
0 x x x	Inverted output BP40 appears on BP41 (BP41=NOT BP40)			
1 x x x	Non-inverted output BP40 appears on BP41 (BP41= BP40)			

2.5.4 Timer/Counter in 16-bit Mode

In 16-bit mode, the Timer 0 and Timer 1 are cascaded thus forming a 16-bit counter (see figure 39) whereby, irrespective of the state of the Timer 0 interrupt mask bit (T0IM), the Timer 1 counts both Timer 0 overflow and compare interrupt events. These are generated according to the state of the Timer 0 Mode Register as described in the T0MO table. The comparators are also cascaded so that when both the Timer 0 and Timer 1 match their respective compare registers, the Timer 1 generates both an output signal and a compare interrupt (if unmasked).

In measurement modes, only the Timer 0 capture register is loaded with the Timer 0 contents on an end-of-measurement event. The Timer 1 capture register operates solely as a shadow register. There is no 16-bit capture operation, so the user program must check if the Timer 1 has incremented between reading the lower and higher byte. Likewise, there is no automatic suppression of spurious interrupts which could conceivably be generated between writing the Timer 0 and Timer 1 compare registers.

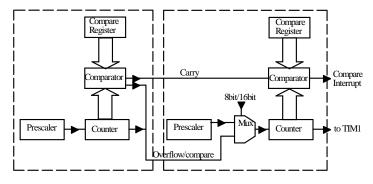


Figure 39. 16-bit mode

2.6 Buzzer Module

The buzzer is a 4 stage frequency divider which divides the SUBCL and depending on the state of the Buzzer Control Register (BZCR) can output one of four frequencies. An external piezo or buzzer can be driven by the complementary buzzer outputs (BUZ and NBUZ) which are directed to Port 4 (BP42 and BP43) under control of the Timer/Counter I/O Register (TCIOR) as shown in figure 23. When the buzzer is switched off, both of the buzzer outputs take up the same logical state. This is controlled by the BZOP bit of the BZCR.

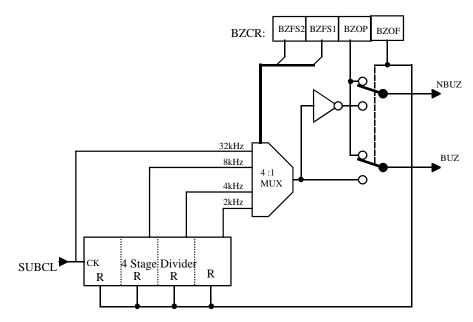


Figure 40. Buzzer module

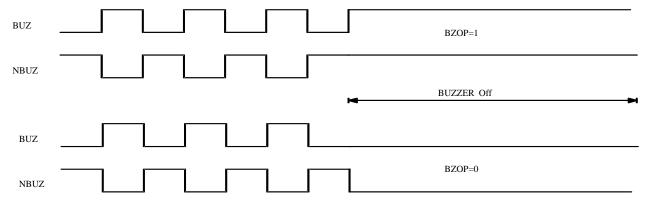


Figure 41. Buzzer waveform

Buzzer Control Register (BZCR)

Subport address (indirect access): 'A'hex

	Bit 3	Bit 2	Bit I	Bit 0	D / 1 4441
BZCR	BZFS2	BZFS1	BZOP	BZOF	Reset value: 1111b

BZFS2, BZFS2 – Buzzer Frequency Select code

BZOP – Buzzer Output Stop State

BZOF – Buzzer off/on

Table 20. Buzzer Control Register (BZCR)

	Timer/Counter Subport Pointer (TCX) Address: '9'hex				
Code	Function				
3 2 1 0					
x x x 0	Buzzer on				
x x x 1	Buzzer off				
x x 0 x	Buzzer output stop state: BP42 = BP43 = low				
x x 1 x	Buzzer output stop state: BP42 = BP43 = high				
0 0 x x	Buzzer frequency: 32 kHz				
0 1 x x	Buzzer frequency: 8 kHz				
1 0 x x	Buzzer frequency: 4 kHz				
1 1 x x	Buzzer frequency: 2 kHz				

2.7 Emulation

For emulation all MARC4 controllers have a special emulation mode. It is activated by setting the TE pin to logic HIGH level after reset. In this mode the internal CPU core is inactive and the I/O bus is available via port 0 and port 1 to allow the emulator the access to the on-chip peripherals. The emulator contains a special emulation

CPU with a MARC4 core and additional breakpoint logic and takes over the core function. The basic function of the emulator is to evaluate the customer's program and hardware in real time. For more informations about emulation see "Emulator Manual".

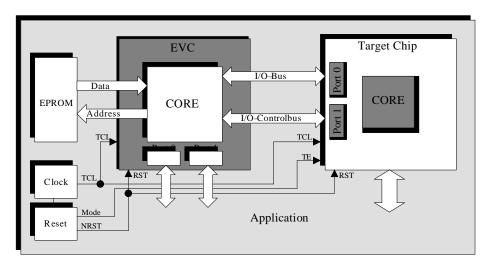


Figure 42. Emulation

3 Electrical Characteristics

3.1 Absolute Maximum Ratings

Voltages are given relative to VSS.

Parameters	Symbol	Value	Unit
Supply voltage	V_{DD}	-0.3 to + 6.5	V
Input voltage (on any pin)	V _{IN}	$V_{SS} - 0.3 \le V_{IN} \le V_{DD} + 0.3$	V
Output short circuit duration	t _{short}	indefinite	sec
Operating temperature range	T _{amb}	-40 to +85	°C
Storage temperature range	T _{stg}	-40 to +130	°C
Thermal resistance (DIP40)	R _{thJA}	110	K/W
Soldering temperature (t ≤ 10 sec)	T _{sd}	260	°C

Stresses greater than those listed under absolute maximum ratings may cause permanent damage to the device. This is a stress rating only and functional operational the device at any condition above those indicated in the operational section of these specification is not implied. Exposure to absolute maximum rating condition for an extended period may affect device reliability. All inputs

and outputs are protected against high electrostatic voltages or electric fields. However, precautions to minimize built-up of electrostatic charges during handling are recommended. Reliability of operation is enhanced if unused inputs are connected to an appropriate logic voltage level (e.g. V_{DD}).

3.2 DC Operating Characteristics

Supply voltage $V_{DD} = 5 \text{ V}$, $V_{SS} = 0 \text{ V}$, $T_{amb} = 25^{\circ}\text{C}$ unless otherwise specified.

Parameters	Test Conditions / Pins	Symbol	Min.	Тур.	Max.	Unit
Power supply						
Active current (CPU active) ext. RC 2.2 MΩ	$\begin{aligned} f_{SYSCL} &= 5 \text{ MHz} \\ V_{DD} &= 2.4 \text{ V} \\ V_{DD} &= 5.0 \text{ V} \\ V_{DD} &= 6.2 \text{ V} \end{aligned}$	I _{DD}		0.35 1.0 1.25		mA mA mA
Power down current (CPU sleep, RC oscillator active)	$V_{DD} = 2.4 \text{ V } f_{SYSCL} = f_{SYSCL} = V_{DD} = 6.2 \text{ V } f_{SYSCL} = $	I _{PD}				mA mA mA mA
Sleep current (CPU sleep, RC oscillator inactive)	$V_{DD} = 2.4 \text{ V}$ $V_{DD} = 6.2 \text{ V}$	I_{Sleep}	1 1	2 3	4 6	μA μA

Parameters	Test Conditions / Pins	Symbol	Min.	Тур.	Max.	Unit		
Power-on reset threshold voltage: Note x figure xx								
POR threshold voltage		V _{POR}		2.1	2.4	V		
POR hysteresis		V _{POR}	0.5			V		
Schmitt-trigger input voltag	Schmitt-trigger input voltage:							
Negative-going threshold voltage	$V_{DD} = 2.4 \text{ to } 6.2 \text{ V}$	V _{T-}	V _{SS}		0.3*V _{DD}	V		
Positive-going threshold voltage	$V_{DD} = 2.4 \text{ to } 6.2 \text{ V}$	V_{T+}	0.7*V _{DD}		V _{DD}	V		
Hysteresis (VT+ – VT–)	$V_{DD} = 2.4 \text{ to } 6.2 \text{ V}$	V_{H}		0.1*V _{DD}				

Input Pins: NRST and TE

Parameters	Test Conditions / Pins	Symbol	Min.	Тур.	Max.	Unit	
Input voltage LOW	$V_{DD} = 2.4 \text{ to } 6.2 \text{ V}$	V_{IL}	V_{SS}		0.2*V _{DD}	V	
Input voltage HIGH	$V_{DD} = 2.4 \text{ to } 6.2 \text{ V}$	V _{IH}	0.8* V _{DD}		V_{DD}	V	
Input NRST with pull-up re	Input NRST with pull-up resistor						
Input LOW current	$V_{DD} = 2.4 \text{ V}, V_{IL} = V_{SS}$	I_{IL}	-100	-125	-150	μΑ	
	$V_{DD} = 5.0 \text{ V}$		-250	-320	-400	μΑ	
Input TE with pull-down resistor							
Input HIGH current	$V_{DD} = 2.4 \text{ V}, \ V_{IH} = V_{DD}$	I_{IH}	15	30	50	μΑ	
	$V_{DD} = 5.0 \text{ V}$		200	260	300	μΑ	

Bidirectional Port BP00...BP03, BP10...BP13 and TIM1

Parameters	Test Conditions / Pins	Symbol	Min.	Тур.	Max.	Unit
Input voltage LOW	$V_{DD} = 2.4 \text{ to } 6.2 \text{ V}$	V_{IL}	V _{SS}		0.2*V _{DD}	V
Input voltage HIGH	$V_{DD} = 2.4 \text{ to } 6.2 \text{ V}$	V _{IH}	0.8* V _{DD}		V_{DD}	V
Input LOW current	$V_{DD} = 2.4 \text{ V}, V_{IL} = V_{SS}$	I_{IL}	-1.0	-1.5	-2.5	μΑ
	$V_{DD} = 5.0 \text{ V}$		-6.0	-8.5	-12	μΑ
Input HIGH current	$V_{DD} = 2.4 \text{ V}, V_{IH} = V_{DD}$	I_{IH}	1.0	1.3	3.0	μΑ
	$V_{DD} = 5.0 \text{ V}$		4.0		10	μΑ
Output LOW current	$V_{DD} = 2.4 \text{ V},$	I_{OL}	1.4	1.8	2.2	mA
	$V_{OL} = 0.2*V_{DD}$		6.0	7.0	8.0	mA
	$V_{DD} = 5.0 \text{ V}$					
Output HIGH current	$V_{DD} = 2.4 \text{ V}$	I _{OH}	-1.0	-1.3	-1.6	mA
	$V_{OH} = 0.8 * V_{DD}$					
	$V_{DD} = 5.0 \text{ V}$		-3.0	-4.0	-5.0	mA

Bidirectional Port BP50...BP53, BP70...BP73

Parameters	Test Conditions / Pins	Symbol	Min.	Тур.	Max.	Unit
Input voltage LOW	$V_{DD} = 2.4 \text{ to } 6.2 \text{ V}$	V _{IL}	V _{SS}		0.2*V _{DD}	V
Input voltage HIGH	$V_{DD} = 2.4 \text{ to } 6.2 \text{ V}$	V _{IH}	0.8* V _{DD}		V_{DD}	V
Input LOW current	$V_{DD} = 2.4 \text{ V}, V_{IL} = V_{SS}$	I_{IL}	-1.0	-1.5	-2.5	μΑ
	$V_{DD} = 5.0 \text{ V}$		-6.0	-8.5	-12	μΑ
Input HIGH current	$V_{DD} = 2.4 \text{ V}, V_{IH} = V_{DD}$	I_{IH}	1.0	1.3	3.0	μΑ
	$V_{DD} = 5.0 \text{ V}$		4.0		10.0	μΑ
Output LOW current	$V_{DD} = 2.4 \text{ V}$	I _{OL}	1.4	1.8	2.2	mA
	$V_{OL} = 0.2 * V_{DD}$					
	$V_{DD} = 5.0 \text{ V}$		6.0	7.0	8.0	mA
Output HIGH current	$V_{DD} = 2.4 \text{ V}$	I _{OH}	-1.0	-1.3	-1.6	mA
	$V_{OH} = 0.8 * V_{DD}$					
	$V_{DD} = 5.0 \text{ V}$		-3.0	-4.0	-5.0	mA

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Bidirectional Port BPA0...BPA3, BPB0...BPB3

Parameters	Test Conditions / Pins	Symbol	Min.	Тур.	Max.	Unit
Input voltage LOW	$V_{DD} = 2.4 \text{ to } 6.2 \text{ V}$	V_{IL}	V_{SS}		0.2*V _{DD}	V
Input voltage HIGH	$V_{DD} = 2.4 \text{ to } 6.2 \text{ V}$	V_{IH}	0.8* V _{DD}		V_{DD}	V
Input LOW current (static pull-up)	$V_{DD} = 2.4 \text{ V}, V_{IL} = V_{SS}$ $V_{DD} = 5.0 \text{ V}$	I_{IL}	-20 -120	-27 -160	-40 -200	μΑ μΑ
Input LOW current (dynamic pull-up)	$V_{DD} = 2.4 \text{ V}, V_{IL} = V_{SS}$ $V_{DD} = 5.0 \text{ V}$	I_{IL}	-1.0 -6.0	-1.5 -8.5	-2.5 -12.0	μΑ μΑ
Input HIGH current (dynamic pull-down)	$V_{DD} = 2.4 \text{ V}, V_{IH} = V_{DD}$ $V_{DD} = 5.0 \text{ V}$	I _{IH}	1.0 4.0	1.3	3.0 10.0	μΑ μΑ
Output LOW current	$V_{DD} = 2.4 \text{ V}$ $V_{OL} = 0.2*V_{DD}$ $V_{DD} = 5.0 \text{ V}$	I _{OL}	1.4 6.0	1.8 7.0	2.2 8.0	mA mA
Output HIGH current $V_{DD} = 2.4 \text{ V}$ $V_{OH} = 0.8*V_{DD}$ $V_{DD} = 5.0 \text{ V}$		I _{OH}	-1.0 -3.0	-1.3 -4.0	-1.6 -5.0	mA mA

Bidirectional Port BP40 and BP41

Parameters	Test Conditions / Pins	Symbol	Min.	Тур.	Max.	Unit
Input voltage LOW	$V_{DD} = 2.4 \text{ to } 6.2 \text{ V}$	V_{IL}	V _{SS}		0.2*V _{DD}	V
Input voltage HIGH	$V_{DD} = 2.4 \text{ to } 6.2 \text{ V}$	V _{IH}	0.8* V _{DD}		V_{DD}	V
Input LOW current	$V_{DD} = 2.4 \text{ V}, V_{IL} = V_{SS}$	I_{IL}	-1.0	-1.5	-2.5	μΑ
(dynamic pull-up)	$V_{DD} = 5.0 \text{ V}$		-6.0	-8.5	-12.0	μΑ
Input HIGH current	$V_{DD} = 2.4 \text{ V}, V_{IH} = V_{DD}$	I_{IH}	1.0	1.3	3.0	μΑ
(dynamic pull-down)	$V_{DD} = 5.0 \text{ V}$		4.0		10.0	μΑ
Output LOW current	$V_{DD} = 2.4 \text{ V}$	I_{OL}	1.4	1.8	2.2	mA
	$V_{OL} = 0.2*V_{DD}$					
	$V_{DD} = 5.0 \text{ V}$		6.0	7.0	8.0	mA
Output HIGH current $V_{DD} = 2.4 \text{ V}$		I _{OH}	-1.0	-1.3	-1.6	mA
	$V_{OH} = 0.8 * V_{DD}$					
	$V_{DD} = 5.0 \text{ V}$		-3.0	-4.0	-5.0	mA

Bidirectional Port BP42 and BP43 (Buzzer)

Rev. A1: 15.08.1995

Parameters	Test Conditions / Pins	Symbol	Min.	Тур.	Max.	Unit
Input voltage LOW	$V_{DD} = 2.4 \text{ to } 6.2 \text{ V}$	$V_{\rm IL}$	V_{SS}		0.2*V _{DD}	V
Input voltage HIGH	$V_{DD} = 2.4 \text{ to } 6.2 \text{ V}$	V _{IH}	0.8* V _{DD}		V_{DD}	V
Input LOW current	$V_{DD} = 2.4 \text{ V}, V_{IL} = V_{SS}$	I_{IL}	-1.0	-1.5	-2.5	μΑ
	$V_{DD} = 5.0 \text{ V}$		-6.0	-8.5	-12.0	μΑ
Input HIGH current	$V_{DD} = 2.4 \text{ V}, V_{IH} = V_{DD}$	I_{IH}	1.0	1.3	3.0	μΑ
	$V_{DD} = 5.0 \text{ V}$		4.0		10.0	μΑ
Output LOW current	$V_{DD} = 2.4 \text{ V}$	I_{OL}	6.0	7.0	8.0	mA
	$V_{OL} = 0.2*V_{DD}$					
	$V_{DD} = 5.0 \text{ V}$		11.0	12.5	14.0	mA
Output HIGH current $V_{DD} = 2.4 \text{ V}$		I _{OH}	-3.5	-4.0	-5.0	mA
	$V_{OH} = 0.8*V_{DD}$					
	$V_{DD} = 5.0 \text{ V}$		-6.0	-7.5	-9.0	mA

Bidirectional Port BP60 and BP61 (INTx, INTy) $\,$

Parameters	Test Conditions / Pins	Symbol	Min.	Тур.	Max.	Unit
Input voltage LOW	$V_{DD} = 2.4 \text{ to } 6.2 \text{ V}$	V_{IL}	V _{SS}		0.2*V _{DD}	V
Input voltage HIGH	$V_{DD} = 2.4 \text{ to } 6.2 \text{ V}$	V_{IH}	0.8* V _{DD}		V_{DD}	V
Input LOW current (static pull-up)	$V_{DD} = 2.4 \text{ V}, V_{IL} = V_{SS}$ $V_{DD} = 5.0 \text{ V}$	I_{IL}	-0.2 -1.4	-0.35 -1.7	-0.65 -2.5	mA mA
Input LOW current (dynamic pull-up)	$V_{DD} = 2.4 \text{ V}, V_{IL} = V_{SS}$ $V_{DD} = 5.0 \text{ V}$	I_{IL}	-1.0 -6.0	-1.5 -8.5	-2.5 -12.0	μΑ μΑ
Input HIGH current (dynamic pull-down)	$V_{DD} = 2.4 \text{ V}, V_{IH} = V_{DD}$ $V_{DD} = 5.0 \text{ V}$	I _{IH}	1.0 4.0	1.3	3.0 10.0	μΑ μΑ
Output LOW current	$V_{DD} = 2.4 \text{ V}$ $V_{OL} = 0.2*V_{DD}$	I _{OL}	1.4 6.0	1.8	2.2 8.0	mA
$V_{DD} = 5.0 \text{ V}$ Dutput HIGH current $V_{DD} = 2.4 \text{ V}$ $V_{OH} = 0.8*V_{DD}$		I _{OH}	-1.0	7.0	-1.6	mA mA
	$V_{DD} = 6.2 \text{ V}$		-3.0	-4.0	-5.0	mA

3.3 AC Characteristics

Supply voltage V_{DD} = 2.4 to 6.2 V, V_{SS} = 0 V, T_{amb} = 25°C unless otherwise specified.

Parameters	Test Conditions / Pins	Symbol	Min.	Тур.	Max.	Unit			
Timer input timing TIM1, BP40 and BP41									
Timer input clock		f _{TIMx}		4	10	MHz			
Timer input LOW time	Rise / fall time < 10 ns	t _{TIL}	50			ns			
Timer input HIGH time	Rise / fall time < 10 ns	t _{TIH}	50			ns			
Interrupt request input tin	ning								
Int. request LOW time	Rise $/ \pm 5$ fall time < 10 ns	t _{IRL}	50			ns			
Int. request HIGH time	Rise / fall time < 10 ns	t _{IRH}	50			ns			
System clock									
SCLIN input clock		f		4		MHz			
SCLOUT output clock		t				ns			
Input HIGH time		t				μs			
Rise time		t				ns			
Fall time		t							
Reset timing									
Power-on reset time	$V_{DD} > V_{POR}$	T _{POR}		200	500	μs			
NRST input LOW time		T _{NRST}	4*SYSCL			μs			
Operation cycle time									
RC oscillator - external re	esistor								
Frequency	Note 1	f _{RCe}	0.8	2.0	4.0	MHz			
Stability	Note 1	$\Delta f/f$			±5	%			
32-kHz oscillator									
Frequency		f_X		32.768		kHz			
Start up time	$V_{DD} = 3.0 \text{ V}$	t_{SQ}		0.5	1	S			
Stability	Note 2	$\Delta f/f$		0.1		ppm			
Integrated input / output		C _{IN}		20		pF			
capacitances		C _{OUT}		20		pF			

Parameters	Test Conditions / Pins	Symbol	Min.	Тур.	Max.	Unit			
External 32-kHz crystal parameters									
Crystal frequency		f_X		32.768		kHz			
Series resistance		RS		30	50	kΩ			
Static capacitance		C0		1.5		pF			
Dynamic capacitance		C1		3		fF			

Crystal Specification



Figure 43. Crystal Equivalent Circuit

4 Pad Layout

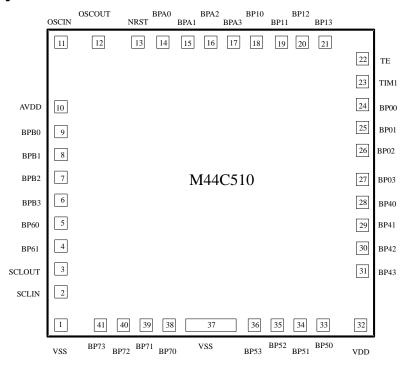


Figure 44. Pad assignments

Table 21. Pad coordinates

Pad No.	Name	X-Coord	Y-Coord	Pad No.	Name	X-Coord	Y-Coord
1	VSS	0.0	0.0	21	BP13	1769.6	1931.6
2	SCLIN	0.0	202.65	22	TE	2103.2	1740.35
3	SCLOUT	0.0	352.65	23	TIM1	2103.2	1560.85
4	BP61	0.0	502.65	24	BP00	2103.2	1410.85
5	BP60	0.0	652.65	25	BP01	2103.2	1260.85
6	BPB3	0.0	802.65	26	BP02	2103.2	1110.85
7	BPB2	0.0	952.65	27	BP03	2103.2	960.85
8	BPB1	0.0	1102.65	28	BP40	2103.2	757.55
9	BPB0	0.0	1252.65	29	BP41	2103.2	607.55
10	AVDD	0.0	1402.65	30	BP42	2103.2	457.55
11	OSCIN	9.8	1931.6	31	BP43	2103.2	307.55
12	OSCOUT	263.05	1931.65	32	VDD	2103.2	0.0
13	NRST	569.65	1931.65	33	BP50	1705.25	0.0
14	BPA0	719.65	1931.6	34	BP51	1555.25	0.0
15	BPA1	869.65	1931.6	35	BP52	1405.25	0.0
16	BPA2	1019.65	1931.6	36	BP53	1255.25	0.0
17	BPA3	1169.65	1931.6	37	VSS	923.55	0.0
18	BP10	1319.6	1931.6	38	BP70	755.55	0.0
19	BP11	1469.6	1931.6	39	BP71	605.55	0.0
20	BP12	1619.6	1931.6	40	BP72	455.55	0.0
				41	BP73	305.55	0.0

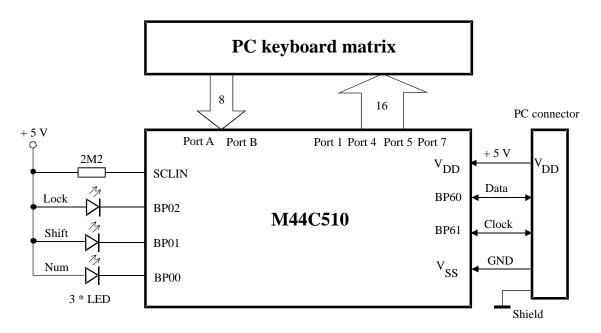


Figure 45. Application example: M44C510 as keyboard controller

TELEFUNKEN Semiconductors

Ozone Depleting Substances Policy Statement

It is the policy of TEMIC TELEFUNKEN microelectronic GmbH to

- 1. Meet all present and future national and international statutory requirements.
- Regularly and continuously improve the performance of our products, processes, distribution and operating systems with respect to their impact on the health and safety of our employees and the public, as well as their impact on the environment.

It is particular concern to control or eliminate releases of those substances into the atmosphere which are known as ozone depleting substances (ODSs).

The Montreal Protocol (1987) and its London Amendments (1990) intend to severely restrict the use of ODSs and forbid their use within the next ten years. Various national and international initiatives are pressing for an earlier ban on these substances.

TEMIC TELEFUNKEN microelectronic GmbH semiconductor division has been able to use its policy of continuous improvements to eliminate the use of ODSs listed in the following documents.

- 1. Annex A, B and list of transitional substances of the Montreal Protocol and the London Amendments respectively
- 2. Class I and II ozone depleting substances in the Clean Air Act Amendments of 1990 by the Environmental Protection Agency (EPA) in the USA
- 3. Council Decision 88/540/EEC and 91/690/EEC Annex A, B and C (transitional substances) respectively.

TEMIC can certify that our semiconductors are not manufactured with ozone depleting substances and do not contain such substances.

We reserve the right to make changes to improve technical design without further notice.

Parameters can vary in different applications. All operating parameters must be validated for each customer application by the customer. Should the buyer use TEMIC products for any unintended or unauthorized application, the buyer shall indemnify TEMIC against all claims, costs, damages, and expenses, arising out of, directly or indirectly, any claim of personal damage, injury or death associated with such unintended or unauthorized use.

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